

Gain-Loss photonic structures:  
formal analogy with quantum-mechanical  
structures with  $\mathcal{PT}$ symmetry/breaking;  
(non-Hermitean theory)

Introduction: Early papers on gain-loss structures  
(without any comment on PT symmetry/breaking)

**Brief history of asymmetric complex grating-assisted coupler:**

***Basic theory:***

L. Poladian, *Phys. Rev. E*, 54, 2963-2975, (1996).

M. Greenberg and M. Orenstein, *Optics Express*, 12, 4013-4018, 2004.

M. Greenberg and M. Orenstein, *Optics Letters*, 29, pp. 451-453, 2004.

M. Greenberg and M. Orenstein, *IEEE J. Quantum Electron.*, 41, 1013-1023, 2005.

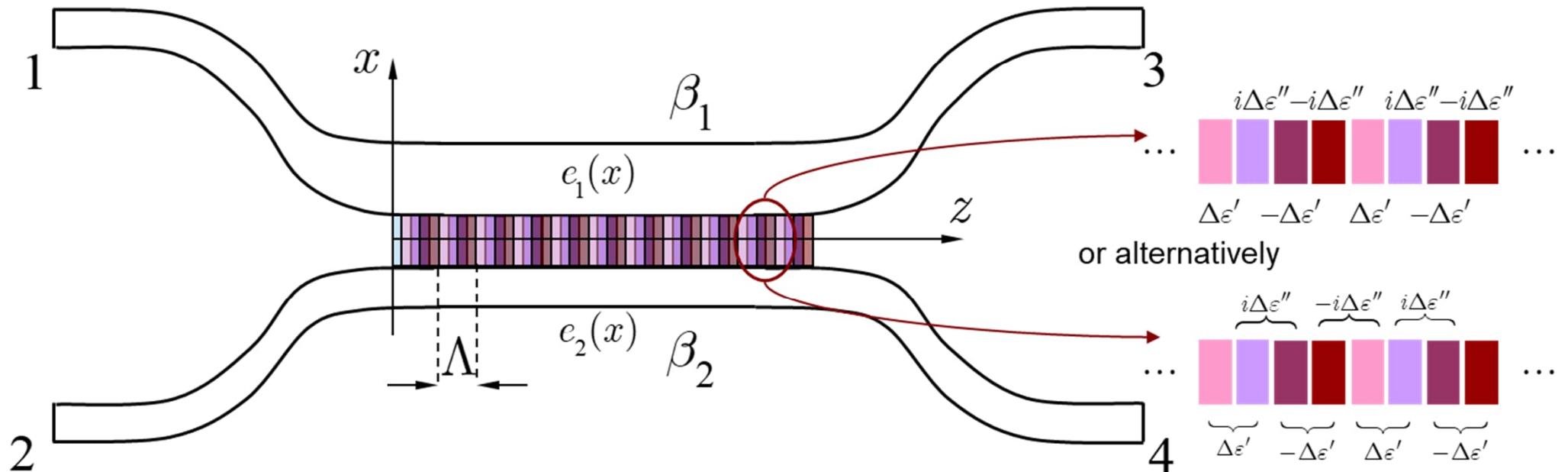
***Application proposals:***

M. Greenberg and M. Orenstein, *Phot. Tech. Lett.*, 17, 1450-1452, 2005. (add mux)

M. Kulishov et al, *Optics Express*, 13, 3567-3578, 2005 (light trapping in ring resonator)

...

# ASYMMETRIC COMPLEX GRATING COUPLER



Grating-assisted directional coupler using **asymmetric complex grating**

$$E_y(x, z) \approx A_1(z)e_1(x)\exp(i\beta_1 z) + A_2(z)e_2(x)\exp(i\beta_2 z);$$

$$\frac{dA_1(z)}{dz} \cong i\kappa_{11}(z)A_1(z) + i\kappa_{12}(z)e^{-i(\beta_1 - \beta_2)z}A_2(z),$$

$$\frac{dA_2(z)}{dz} \cong i\kappa_{21}(z)e^{i(\beta_1 - \beta_2)z}A_1(z) + i\kappa_{22}(z)A_2(z),$$

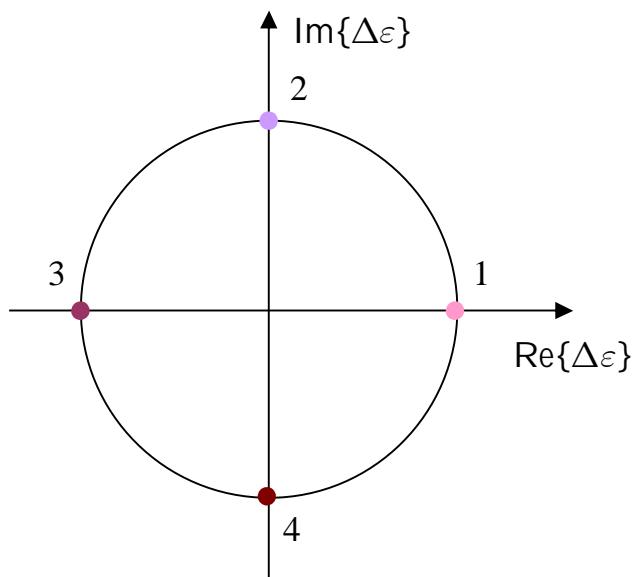
complex, periodic in  $z$

$$\begin{aligned} \kappa_{mn}(z) &= \frac{k_0}{2} \iint_S \Delta\epsilon(x, z)c_m(x)c_n(x)dS \\ &= \kappa_{mn}^{(1)}e^{iKz} + \kappa_{mn}^{(2)}e^{2iKz} + \dots, \quad K = 2\pi/\Lambda \end{aligned}$$

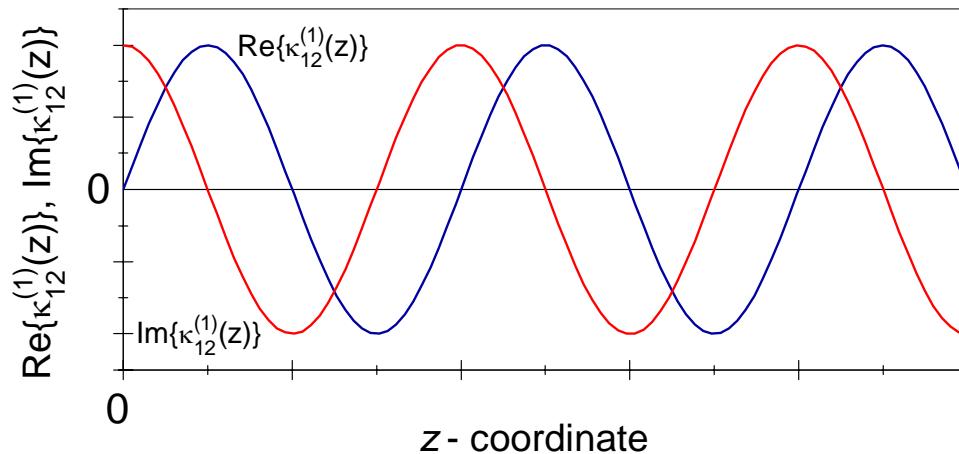
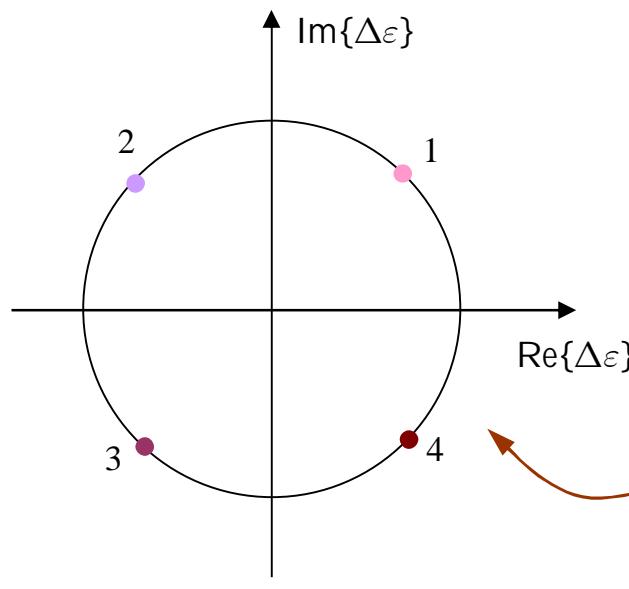
Fourier expansion contains only  
**positive exponentials** (“SSB modulation”)

# ASYMMETRIC COMPLEX GRATING

Complex permittivity perturbation in individual grating segments:



or



The second option seems  
technologically simpler  
since it requires  
only *two different values*  
of  $\Delta\epsilon'$  and  $\Delta\epsilon''$

# SOLUTION OF COUPLED MODE EQUATIONS

Let us consider the following ideal case of the grating at synchronism,

$$\kappa_{11}(z) = 0, \quad \kappa_{12}(z) = \kappa_{12}^{(1)} \exp(iKz),$$

$$\kappa_{21}(z) = 0, \quad \kappa_{22}(z) = 0,$$

$$\Delta\beta = K - (\beta_1 - \beta_2) = 0.$$

Then, the coupled equations read

$$\frac{dA_1(z)}{dz} \cong i\kappa_{12}^{(1)} A_2(z),$$

$$\frac{dA_2(z)}{dz} \cong 0.$$

For  $A_1(0) \neq 0, \quad A_2(0) = 0$

we get the solution

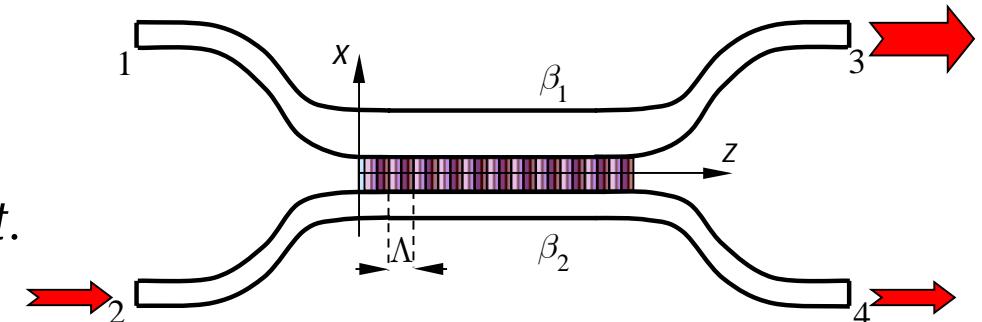
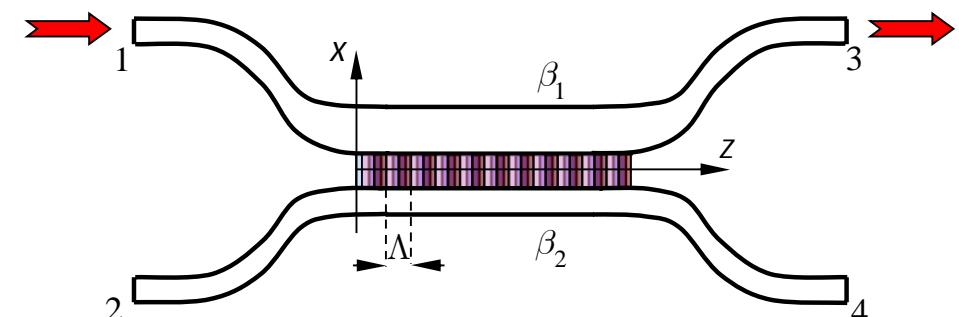
$$A_1(z) = A_1(0) = \text{const.}, \quad P_1(z) = P_1(0),$$

$$A_2(z) = 0, \quad P_2 = 0.$$

for  $A_1(0) = 0, \quad A_2(0) \neq 0$

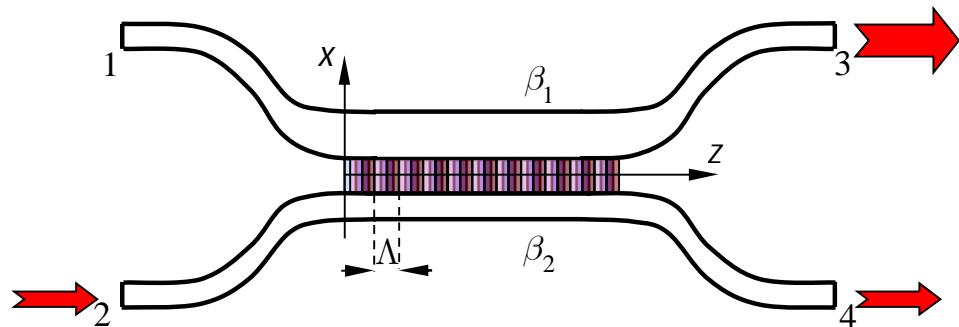
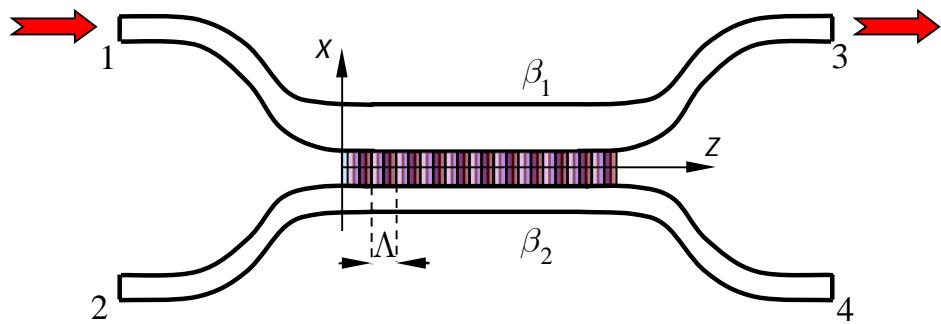
$$A_1(z) = i\kappa_{12}^{(1)} A_2(0)z, \quad P_1(z) = \left| \kappa_{12}^{(1)} \right|^2 P_2(0)z^2,$$

$$A_2(z) = A_2(0) = \text{const.}, \quad P_2(z) = P_2(0) = \text{const.}$$

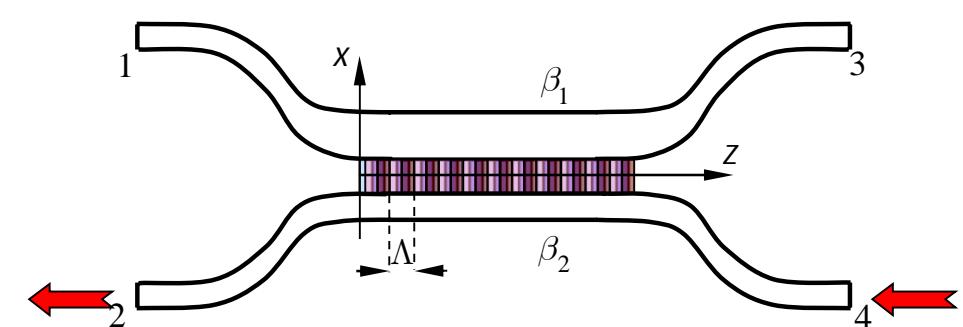
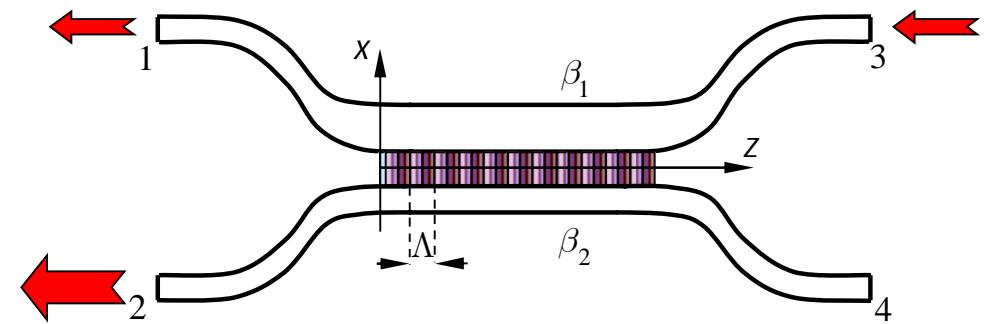


# ACGC IS A RECIPROCAL DEVICE!

Forward propagation



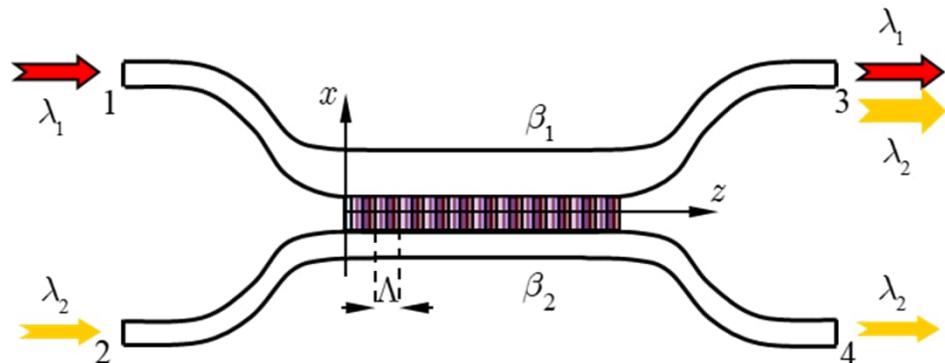
Backward propagation



# STRAIGHTFORWARD ACGC APPLICATIONS

## Wideband ADD multiplexor

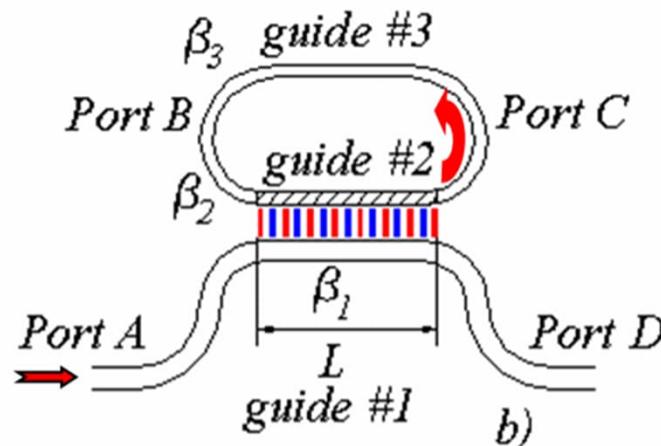
M. Greenberg and M. Orenstein,  
*PTL* 17, 1450-1452, 2005



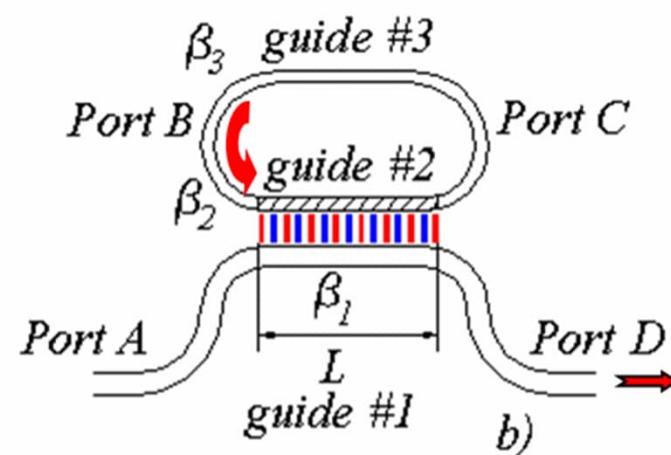
## Light trapping in a ring resonator (a “dynamic memory cell”)

M. Kulishov et al., *OE* 13, 3567-3578, 2005

grating “switched on”:



grating “switched off”:



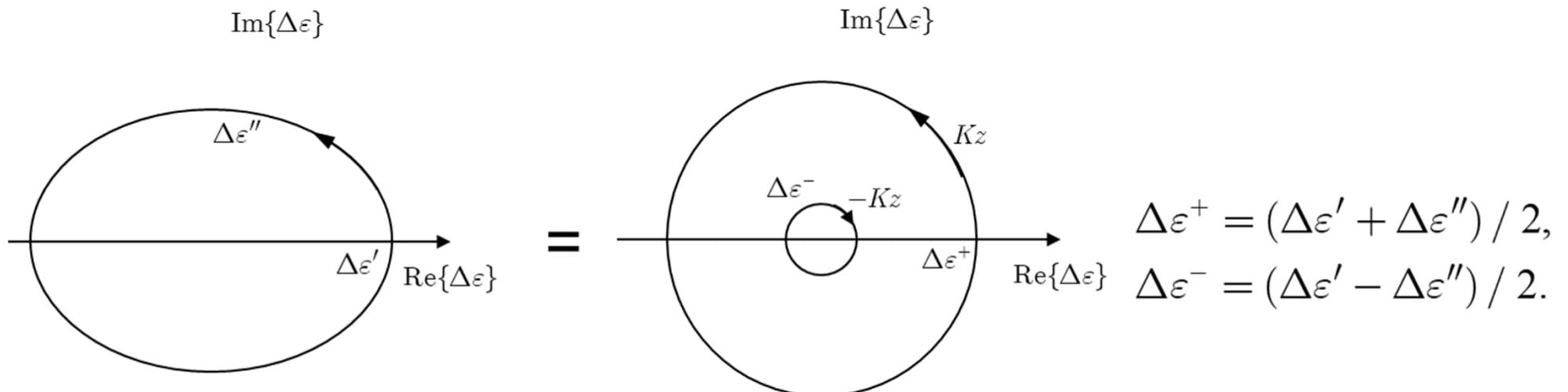
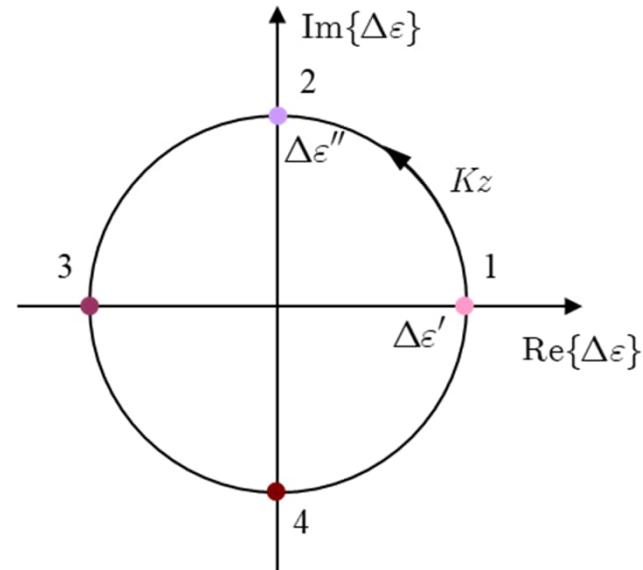
# “SSB” MODULATION

For “unidirectional” behaviour, the condition  $\Delta\varepsilon' = \Delta\varepsilon''$  is of key importance:

$$\dots \begin{array}{ccccccccc} i\Delta\varepsilon'' & -i\Delta\varepsilon'' & i\Delta\varepsilon'' & -i\Delta\varepsilon'' \\ \text{pink} & \text{purple} & \text{dark purple} & \text{red} & \text{pink} & \text{purple} & \text{dark purple} & \text{red} \\ \Delta\varepsilon' & -\Delta\varepsilon' & \Delta\varepsilon' & -\Delta\varepsilon' & \dots \end{array}$$

$$\begin{aligned}\Delta\varepsilon(z) &= \Delta\varepsilon' \cos Kz + i\Delta\varepsilon'' \sin Kz \\ &= \frac{1}{2}(\Delta\varepsilon' + \Delta\varepsilon'')e^{iKz} + \frac{1}{2}(\Delta\varepsilon' - \Delta\varepsilon'')e^{-iKz}\end{aligned}$$

In the case of  $\Delta\varepsilon' \neq \Delta\varepsilon''$  we get

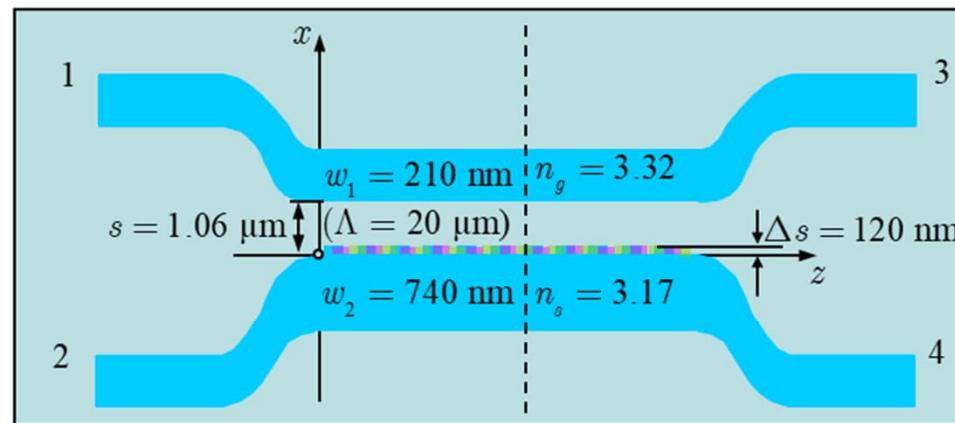


# NUMERICAL MODELLING OF ACGC

**Method:** Bi-directional mode expansion propagation based on harmonic expansion with complex coordinate transformation as a PML (BEXX)

J. Čtyroký, *OQE* **38**, pp. 45-62, 2006; *JLT* **25**, No. 9, pp. 2321-2330, 2007; *JLT* **27**, 2009 (in print)

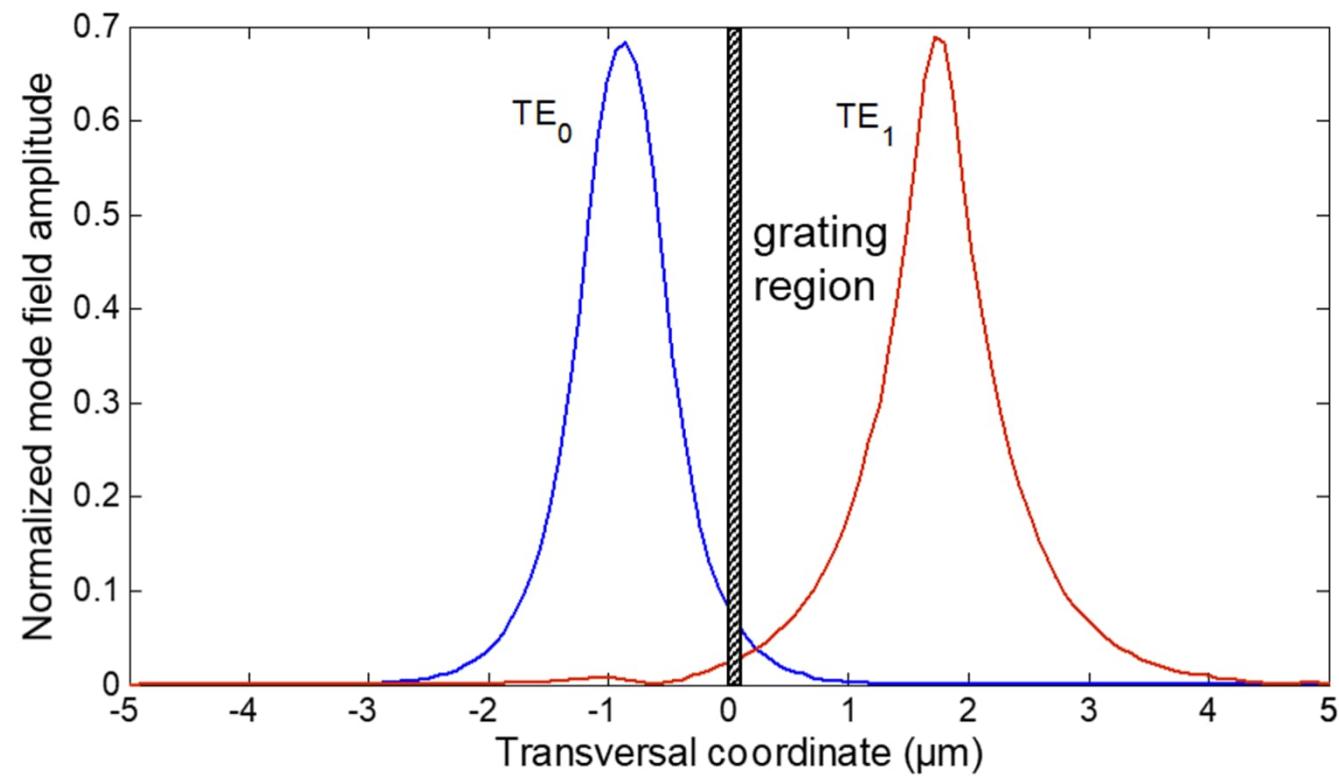
**Basic waveguide structure:** asymmetric directional coupler of the InP/GaInAsP type



**Asymmetric grating:** 24 periods, 4 segments, 5  $\mu\text{m}$  long each  
modulation format: alternative,  
total length of the grating:  $24 \times 4 \times 5 \mu\text{m} = 480 \mu\text{m}$   
central wavelength:  $\lambda = 1.532 \mu\text{m}$   
 $\Delta\epsilon' = \Delta\epsilon'' = 0.48675$  (!!) (gain/loss  $\sim 135 \text{ dB/mm!}$ )

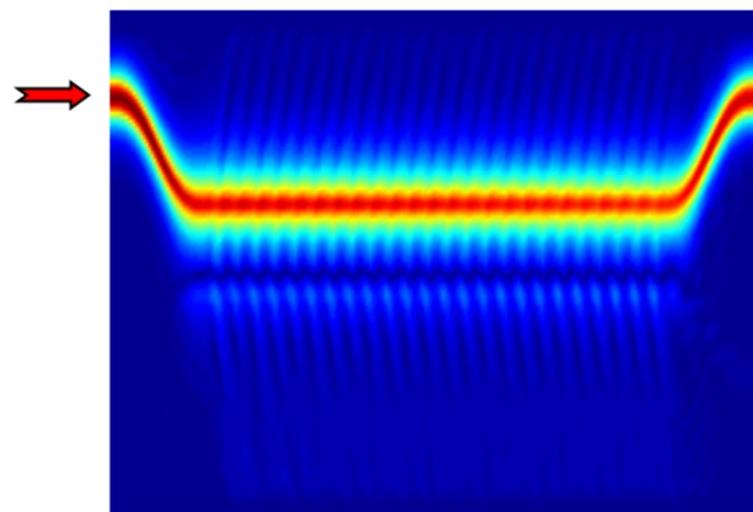
# “STANDARD” ACGC

Mode field distribution in the central part of the coupler

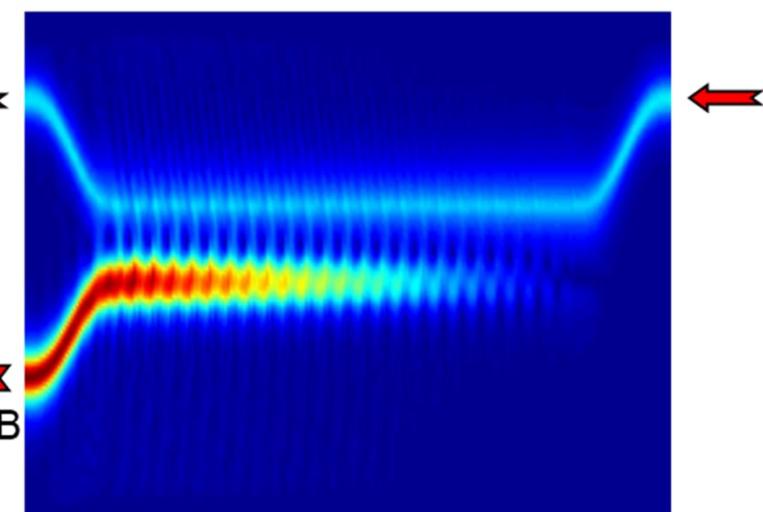


# “STANDARD” ACGC – RESULTS

Forward propagation



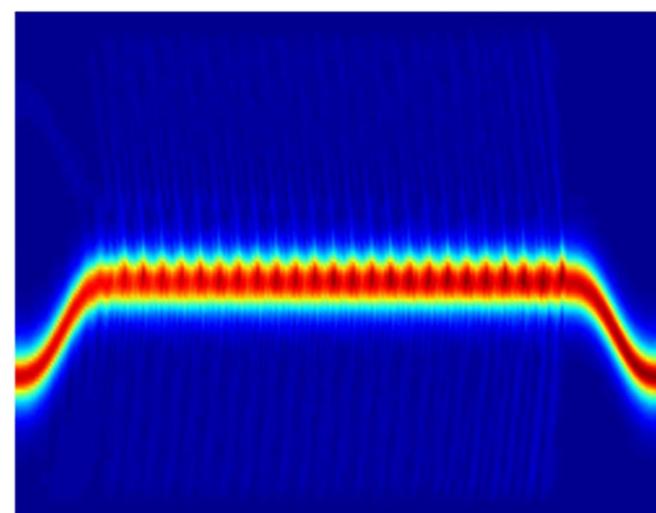
Backward propagation



+9.22 dB  
(gain)

-0.73 dB

+0.19 dB



# FORMAL ANALOGY BETWEEN A PHOTONIC WAVEGUIDE AND A POTENTIAL WELL IN QUANTUM MECHANICS

Eigenmode equation for TE modes  
of a planar waveguide

$$\frac{1}{k_0^2} \frac{d^2 E(x)}{dx^2} + \varepsilon(x) E(x) = N^2 E(x)$$

mode field distribution

$E(x)$   $\Leftrightarrow$   $\psi(x)$  wave function

wave number

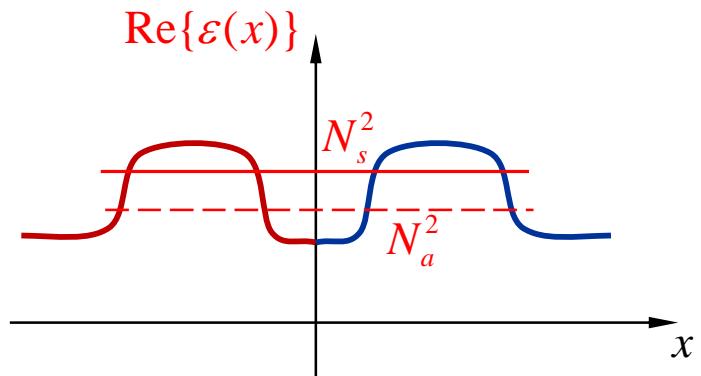
$k_0$   $\Leftrightarrow$   $\frac{\sqrt{2m}}{\hbar}$  mass; Planck constant

relative permittivity profile

$\varepsilon(x)$   $\Leftrightarrow$   $-V(x)$  potential

effective refractive index

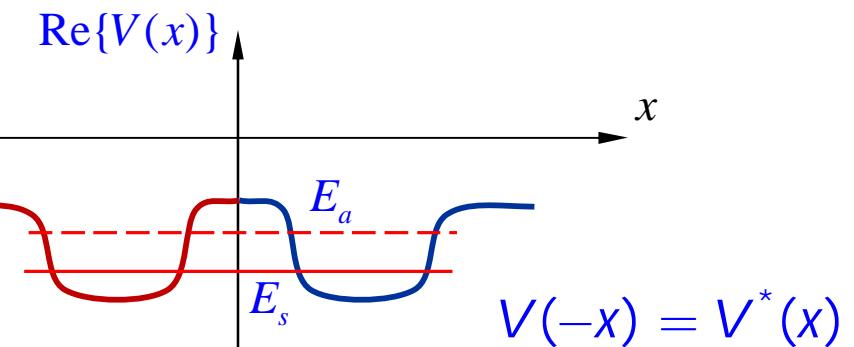
$N^2$   $\Leftrightarrow$   $-E$  particle energy



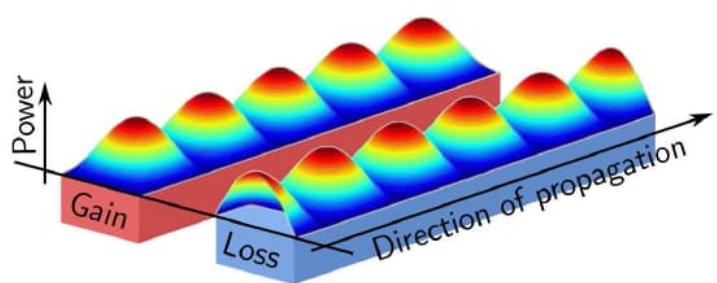
Loss/gain structure:  $\varepsilon(-x) = \varepsilon^*(x)$

Schrödinger equation for a particle in  
a 1D potential well

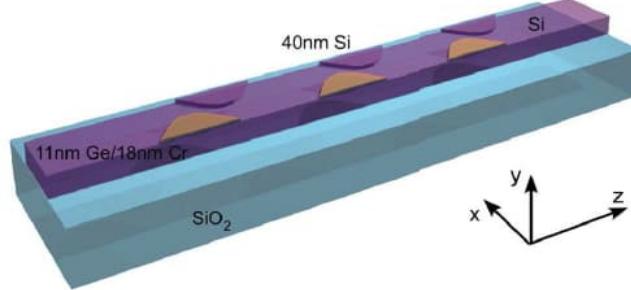
$$-\frac{\hbar^2}{2m} \frac{d^2 \psi(x)}{dx^2} + V(x) \psi(x) = E \psi(x)$$



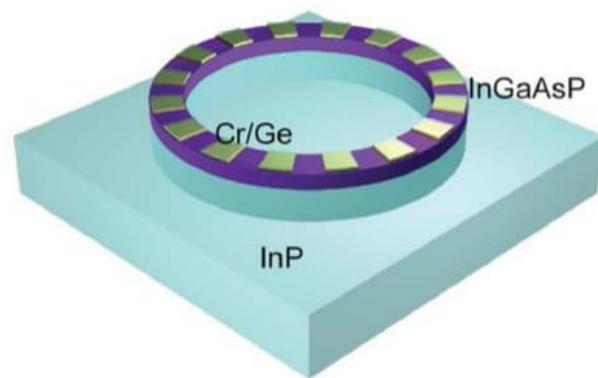
“ $\mathcal{PT}$  symmetry”: complex potential(!),



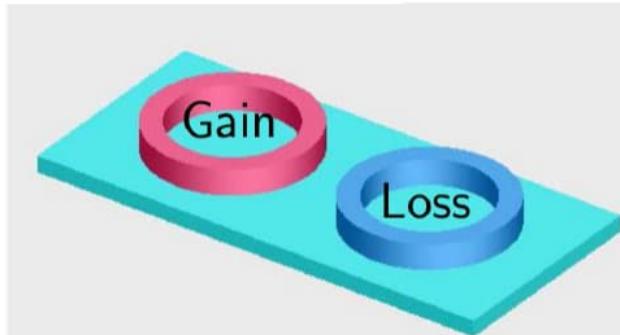
$\mathcal{PT}$ -symmetric  
coupled waveguide [1.34]



$\mathcal{PT}$ -symmetric  
waveguide grating [1.74]



$\mathcal{PT}$ -symmetric  
ring grating [1.61]



$\mathcal{PT}$ -symmetric coupled  
microresonators [1.11, 1.58]

1.34: C. E. Rüter et al., *Nat. Phys.* 6, 192-195 (2010).

1.61: L. Feng et al., *Science* 346, 972-975 (2014).

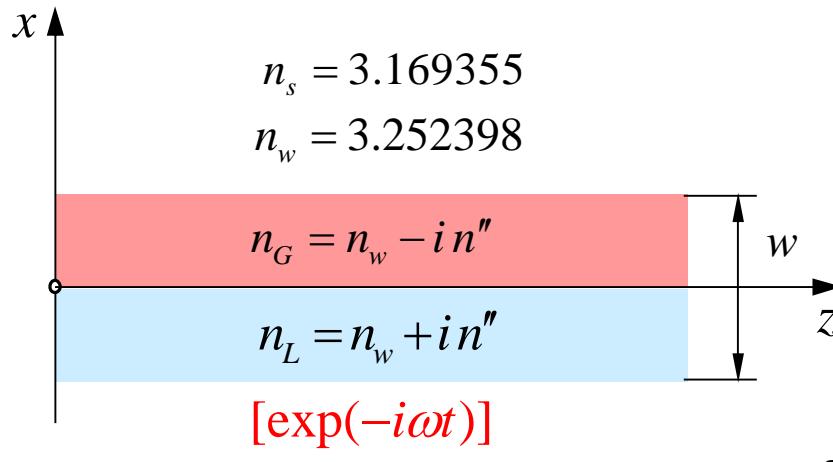
1.74: L. Feng et al., *Science* 333, 729-733 (2011).

1.11: L. Chang et al., *Nat. Photonics* 8, 524-529 (2014).

1.58: H. Hodaei et al., *Science* 346, 975-978 (2014).

# WAVEGUIDE STRUCTURE WITH LOSS/GAIN: HISTORICAL REMARKS

≈ 1995: COST 240 Action: Loss/gain waveguide modelling task by H. P. Nolting (HHI)  
(aimed at benchmarking of BPM methods)

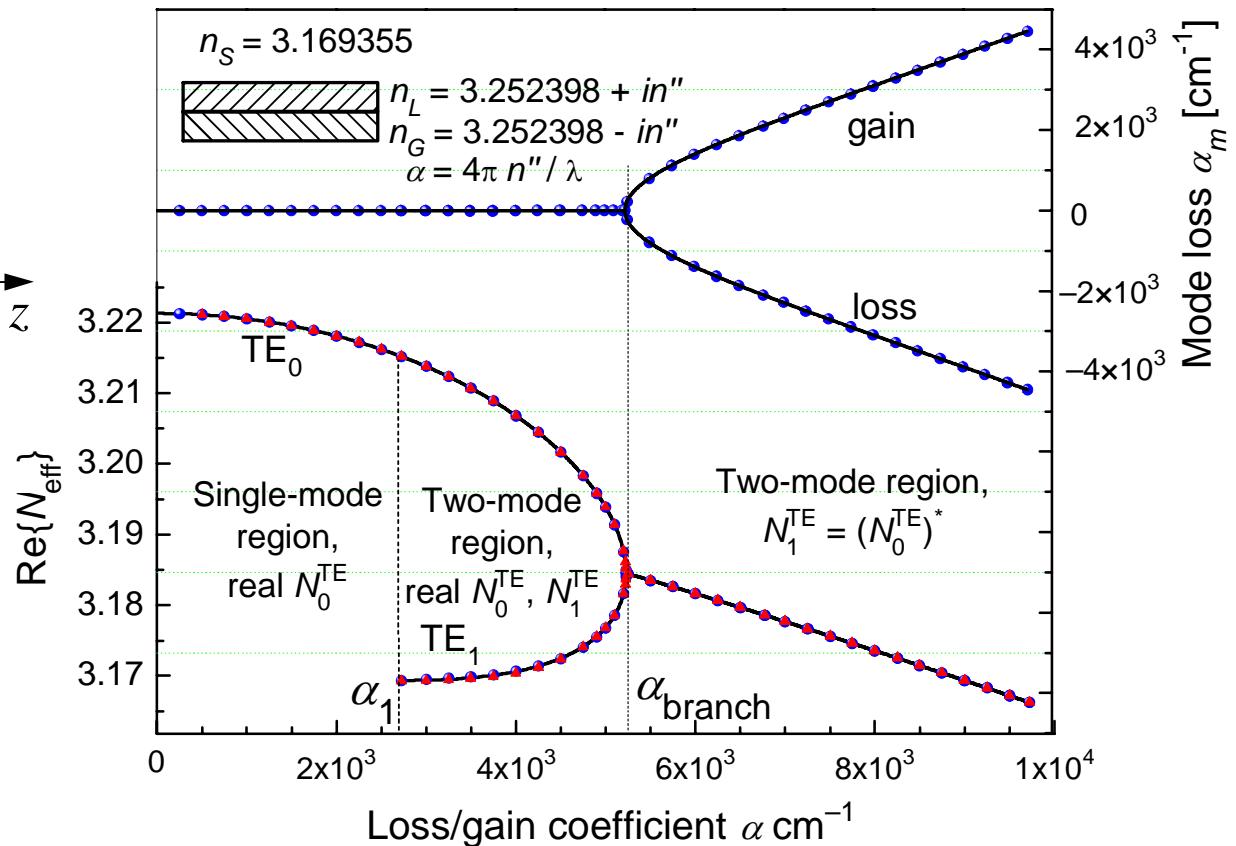


$$w = 1 \mu\text{m},$$

$$n'' = \frac{\alpha}{2k_0} = \frac{\lambda}{4\pi} \alpha \times 10^{-4} \quad [; \mu\text{m}, \text{cm}^{-1}],$$

$$\lambda = 1.55 \mu\text{m},$$

$\alpha$  ... "loss/gain coefficient" [ $\text{cm}^{-1}$ ]



1. H.-P. Nolting, G. Sztefka, J. Čtyroký, "Wave Propagation in a Waveguide with a Balance of Gain and Loss," Integrated Photonics Research '96, Boston, USA, 1996, pp. 76-79.
2. G. Guekos, Ed., Photonic Devices for telecommunications: how to model and measure. Berlin: Springer, 1998, pp. 76-78.

# DISPERSION EQUATION (TE modes)

$$\Phi(N, \alpha) = \gamma_G [\gamma_S \cos(k_0 \gamma_G w) - \gamma_G \sin(k_0 \gamma_G w)] [\gamma_S \sin(k_0 \gamma_L w) + \gamma_L \cos(k_0 \gamma_L w)] + \\ + \gamma_L [\gamma_S \cos(k_0 \gamma_L w) - \gamma_L \sin(k_0 \gamma_L w)] [\gamma_S \sin(k_0 \gamma_G w) + \gamma_G \cos(k_0 \gamma_G w)] = 0$$

$$\gamma_S = \sqrt{N^2 - n_s^2}, \quad \gamma_L = \sqrt{n_L^2 - N^2}, \quad \gamma_G = \sqrt{n_G^2 - N^2}, \quad k_0 = \frac{2\pi}{\lambda}$$

"Exceptional" point:  $\frac{dN}{d\alpha} \rightarrow \infty$ .

$$\text{Since } \Phi(N, \alpha) \equiv 0, \quad \frac{\partial \Phi}{\partial N} \frac{\partial N}{\partial \alpha} + \frac{\partial \Phi}{\partial \alpha} = 0 \quad \Rightarrow \quad \frac{\partial N}{\partial \alpha} = -\frac{\partial \Phi}{\partial \alpha} / \frac{\partial \Phi}{\partial N} \rightarrow \infty \quad \Rightarrow \quad \frac{\partial \Phi}{\partial N} = 0.$$

Exceptional point is given by the simultaneous solution of the following two equations:

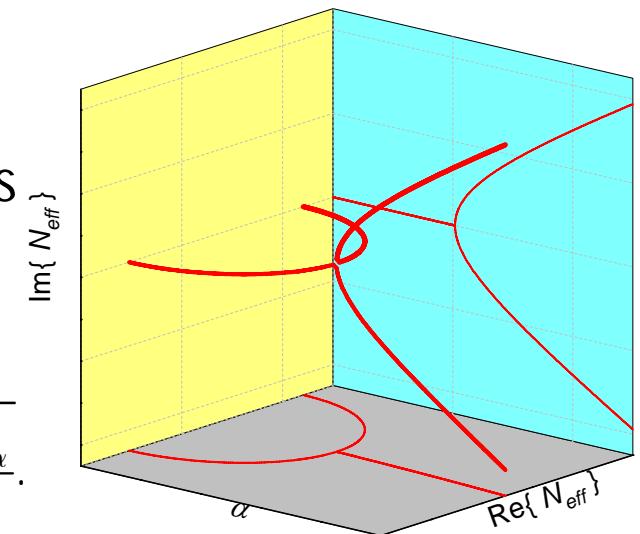
$$\boxed{\Phi(N_B, \alpha_B) = 0, \quad \Phi'_N = \frac{d\Phi(N_B, \alpha_B)}{dN} = 0}$$

Taylor expansion of  $\Phi(N, \alpha)$  in the vicinity of  $N_B, \alpha_B$  sounds

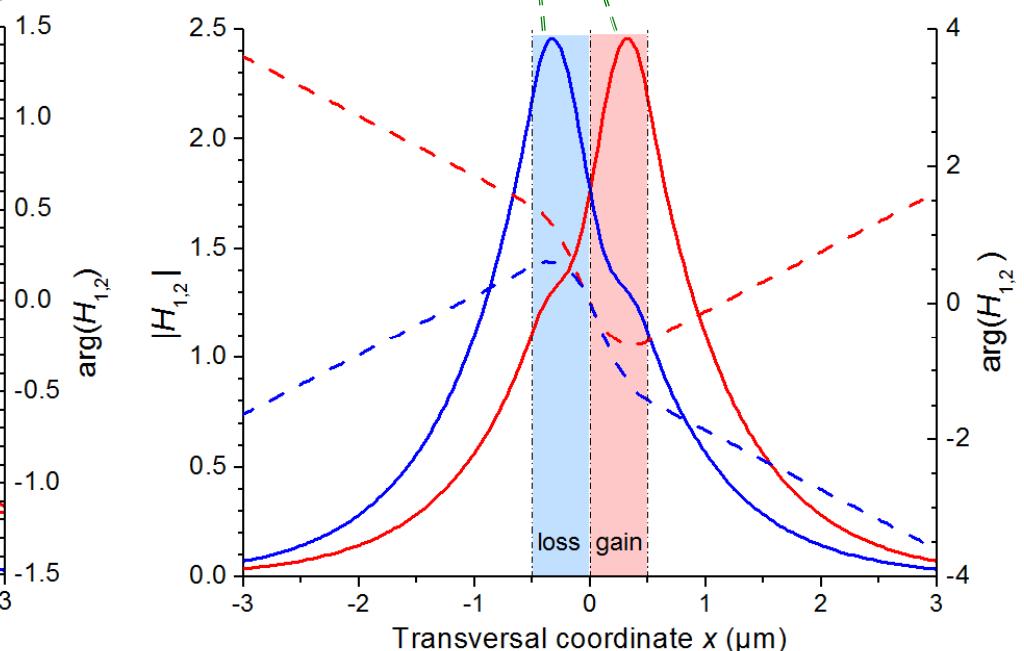
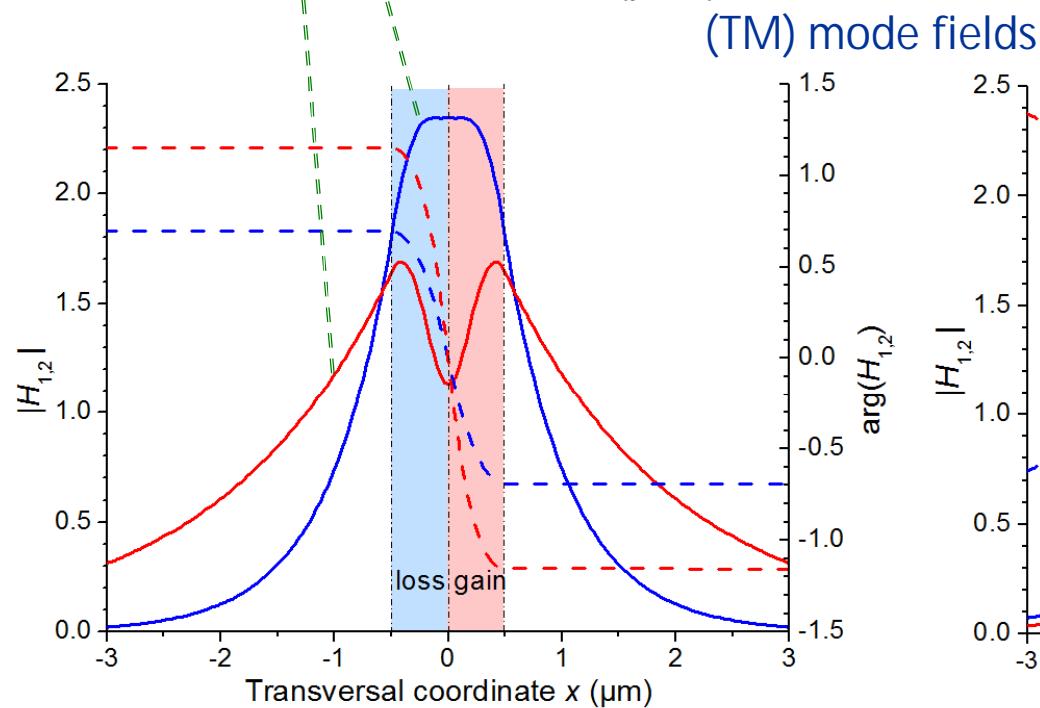
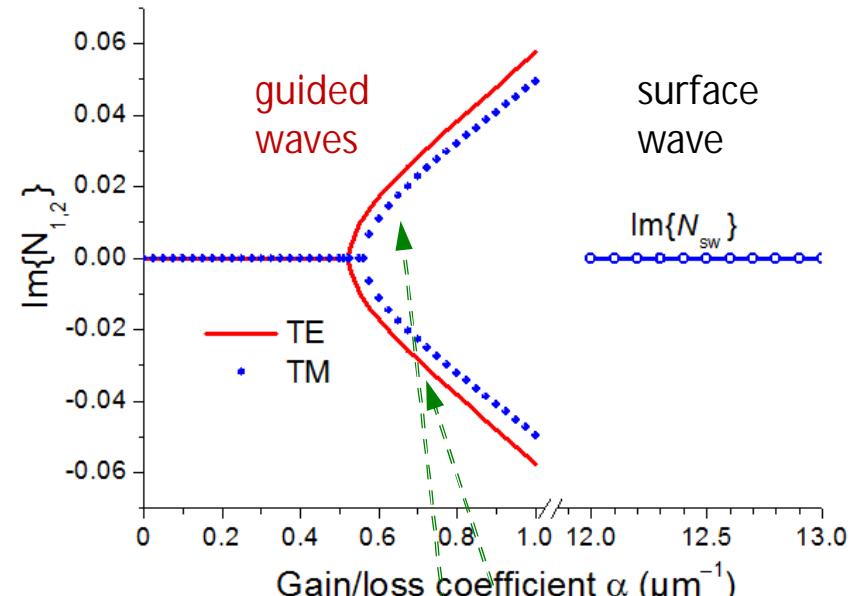
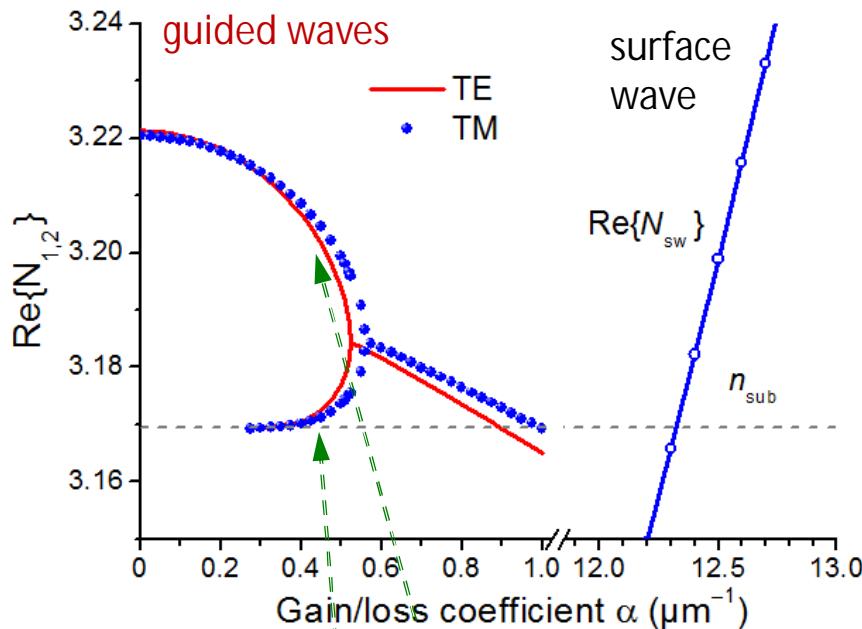
$$\Phi(N, \alpha) \approx \Phi'_a(\alpha - \alpha_B) + \frac{1}{2}\Phi''_N (N - N_B)^2 = 0$$

from which it follows

$$\boxed{N \approx N_B \pm iC \sqrt{(\alpha - \alpha_B)}}, \quad C = \sqrt{\frac{2\Phi'_a}{\Phi''_N}}$$

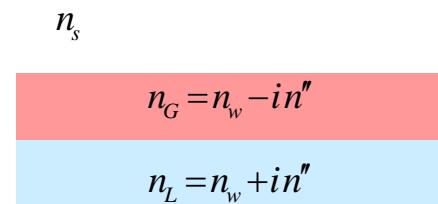
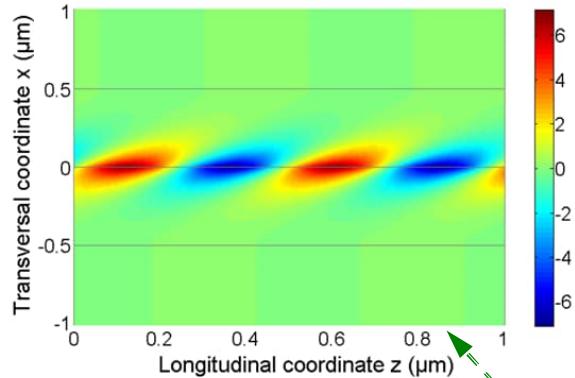


# 2D ANALYSIS (PLANAR WAVEGUIDES)

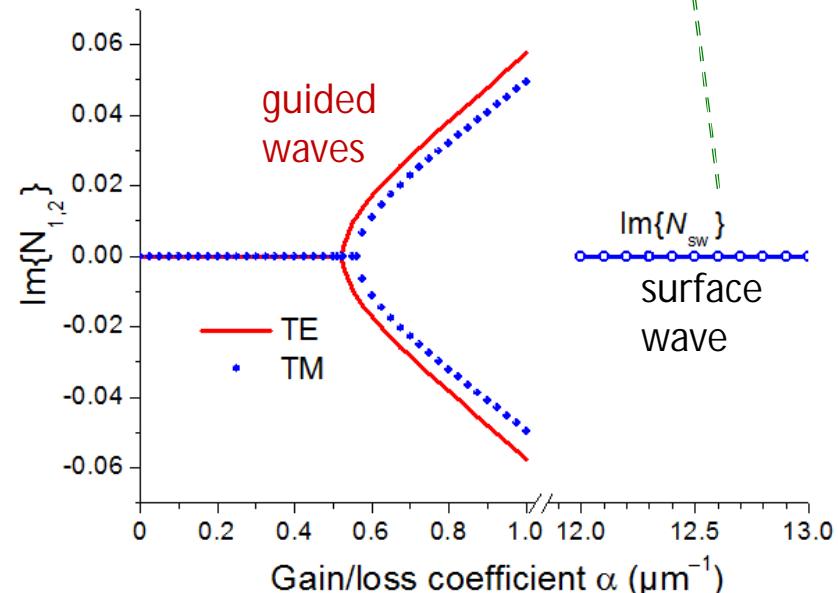
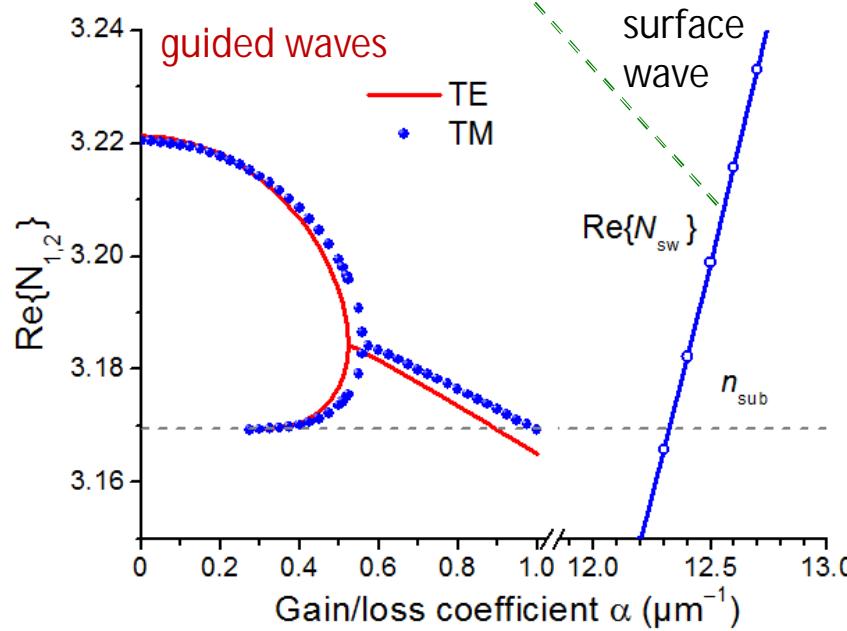
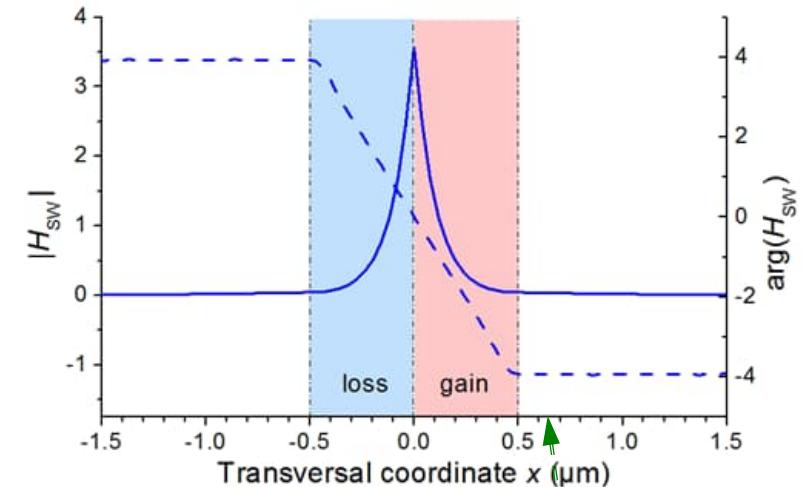


# WAVEGUIDE STRUCTURE WITH LOSS/GAIN: SURFACE WAVE

Non-attenuated TM surface wave



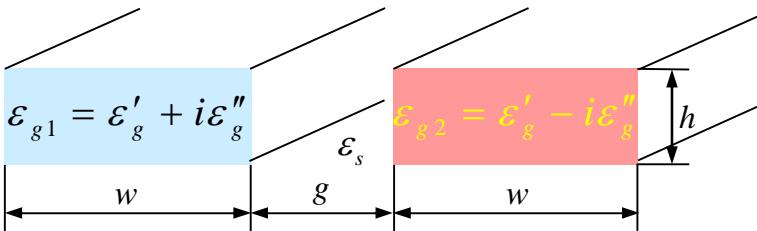
$$N_{SW} = \sqrt{\frac{n_G^2 n_L^2}{n_G^2 + n_L^2}}$$



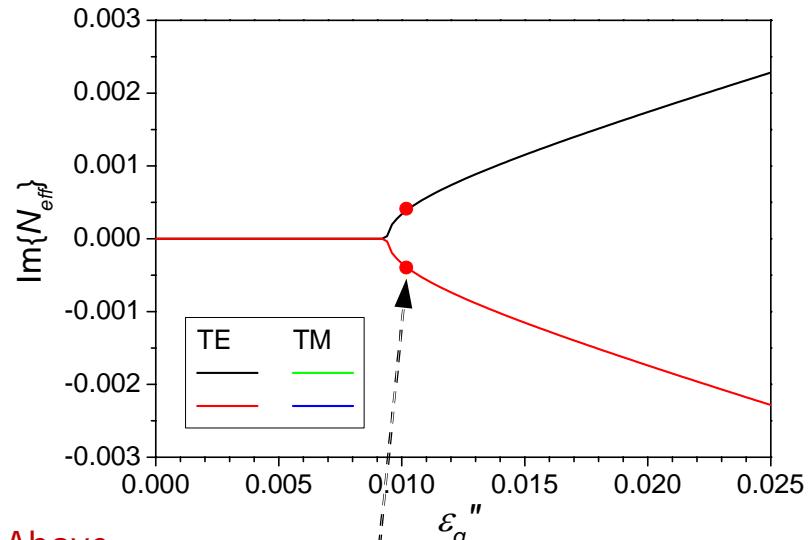
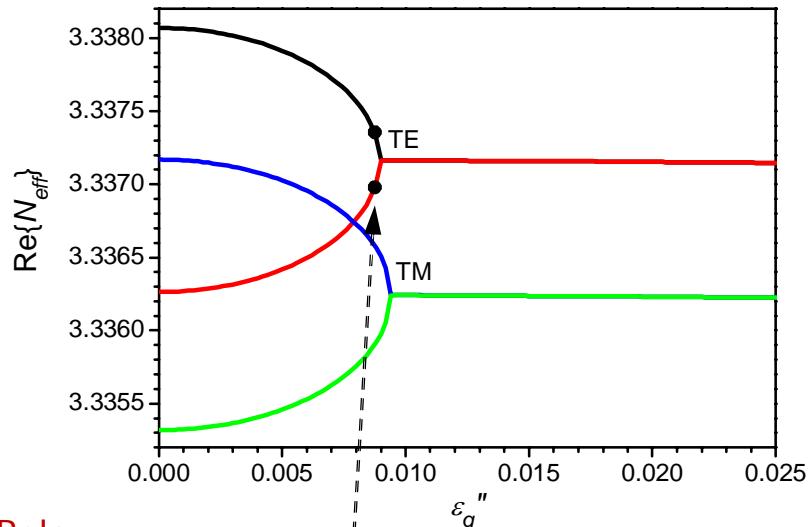
J. Čtyroký et al., "Waveguide structures with antisymmetric gain/loss profile," *Optics Express*, vol. 18, pp. 21585-21593, 2010.

# COUPLED WAVEGUIDES WITH LOSS/GAIN

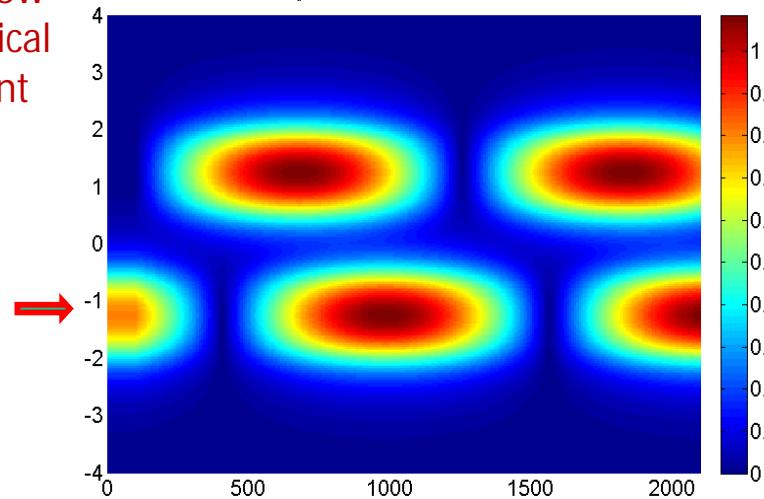
Balanced loss/gain switching:  
 $\varepsilon(-x, y) = \varepsilon^*(x, y)$



$$\begin{aligned} \varepsilon_s &= 10.89, \quad \varepsilon'_g = 11.56 \\ w &= 1.5 \text{ } \mu\text{m}, \quad h = 0.75 \text{ } \mu\text{m}, \\ g &= 1 \text{ } \mu\text{m}, \quad \lambda = 1.55 \text{ } \mu\text{m}. \end{aligned}$$



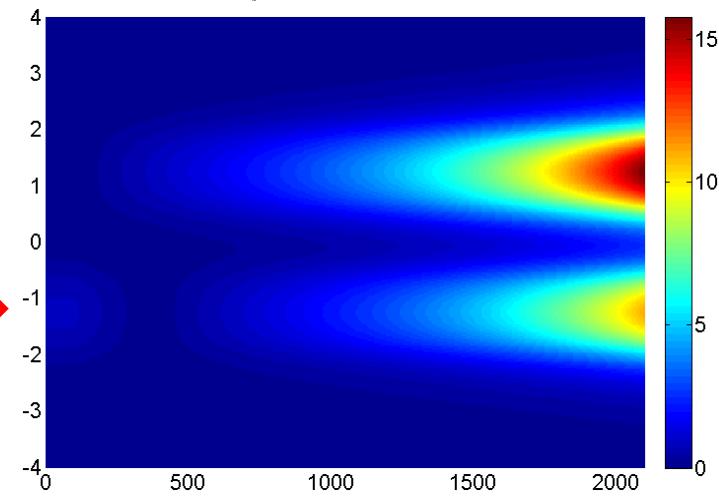
Below critical point



gain channel

loss channel

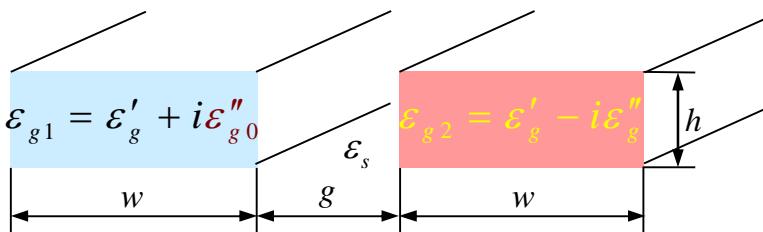
Above critical point



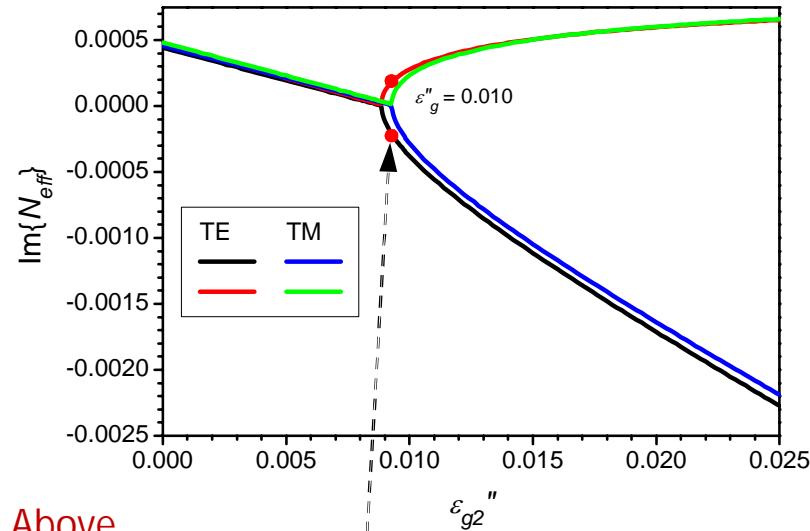
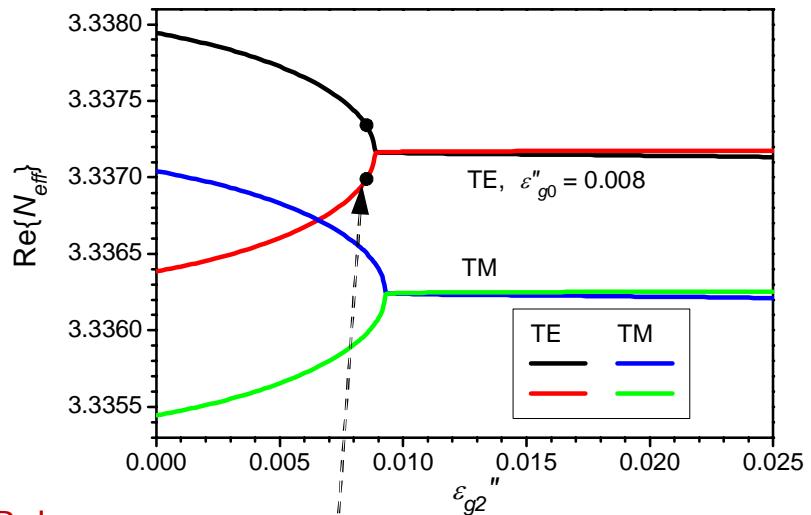
C. E. Rüter et al. "Observation of parity-time symmetry in optics," *Nature Physics*, vol. 6, pp. 192-195, 2010.

# COUPLED WAVEGUIDES WITH LOSS/GAIN

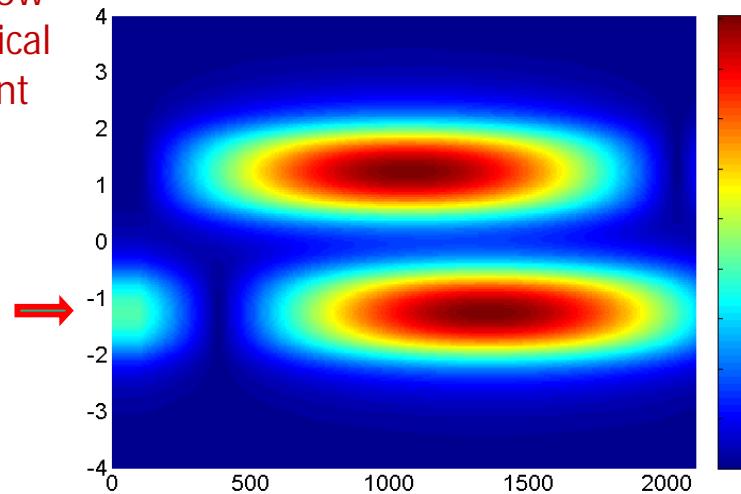
Fixed loss/variable gain switching:  
 $\varepsilon(-x, y) \neq \varepsilon^*(x, y)$



$$\begin{aligned} \varepsilon_s &= 10.89, \quad \varepsilon'_g = 11.56 \\ w &= 1.5 \text{ } \mu\text{m}, \quad h = 0.75 \text{ } \mu\text{m}, \\ g &= 1 \text{ } \mu\text{m}, \quad \lambda = 1.55 \text{ } \mu\text{m}. \end{aligned}$$



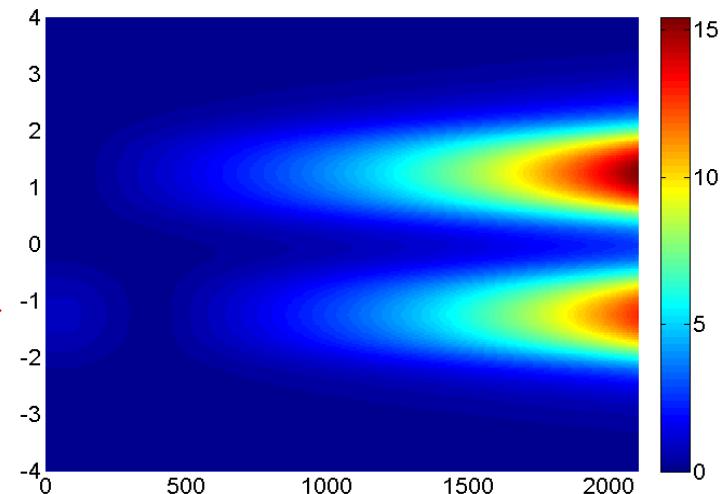
Below  
critical  
point



1.8  
1.6  
1.4  
1.2  
1  
0.8  
0.6  
0.4  
0.2

gain channel  
loss channel

Above  
critical  
point

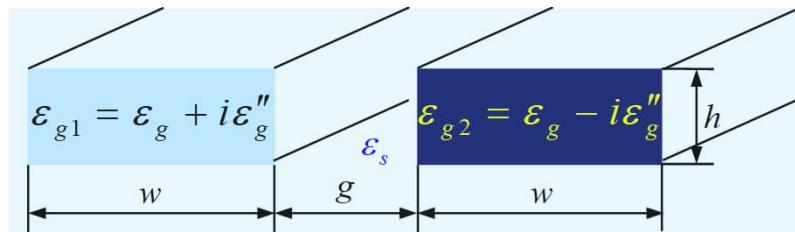


15  
10  
5  
0

# COUPLED WAVEGUIDES WITH LOSS/GAIN

Balanced loss/gain  
with reduced gain:

$$\mathbf{E}(x, y, z) = \mathbf{e}(x, y) \exp(ik_0 N z)$$



$$\varepsilon_s = 10.89 + i\varepsilon''_b, \quad \varepsilon_g = 11.56 + i\varepsilon''_b$$

$$w = 1.5 \text{ } \mu\text{m}, \quad h = 0.75 \text{ } \mu\text{m},$$

$$g = 1 \text{ } \mu\text{m}, \quad \lambda = 1.55 \text{ } \mu\text{m}.$$

Rigorous equation for transversal field components:  $\Delta_{\perp} \mathbf{e}_{\perp} + \nabla_{\perp} [\nabla_{\perp} (\ln \varepsilon) \cdot \mathbf{e}_{\perp}] + k_0^2 (\varepsilon - N^2) \mathbf{e}_{\perp} = \mathbf{0},$

Small uniform permittivity modification:  $\varepsilon_1(x, y) = \varepsilon(x, y) + i\varepsilon''_b, \quad |\varepsilon_b| \ll |\varepsilon|$

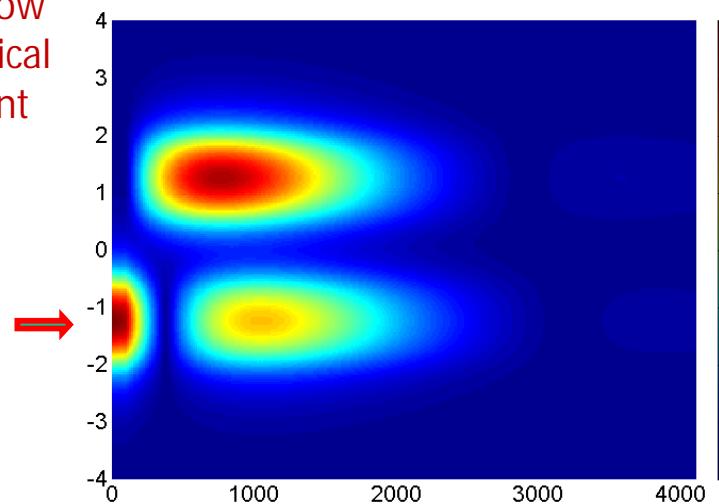
$$\Delta_{\perp} \mathbf{e}_{\perp} + \nabla_{\perp} \left[ \nabla_{\perp} \left( \ln \varepsilon + \frac{i\varepsilon''_b}{\varepsilon} \right) \cdot \mathbf{e}_{\perp} \right] + k_0^2 [\varepsilon + i\varepsilon''_b - (N^2 + i\varepsilon''_b)] \mathbf{e}_{\perp} = \mathbf{0}; \quad \varepsilon \rightarrow \varepsilon + i\varepsilon''_b \Rightarrow N^2 \rightarrow N^2 + i\varepsilon''_b$$

Uniform background loss can be used to reduce the required gain:

$$\varepsilon''_{branch} = \pm 0.009$$

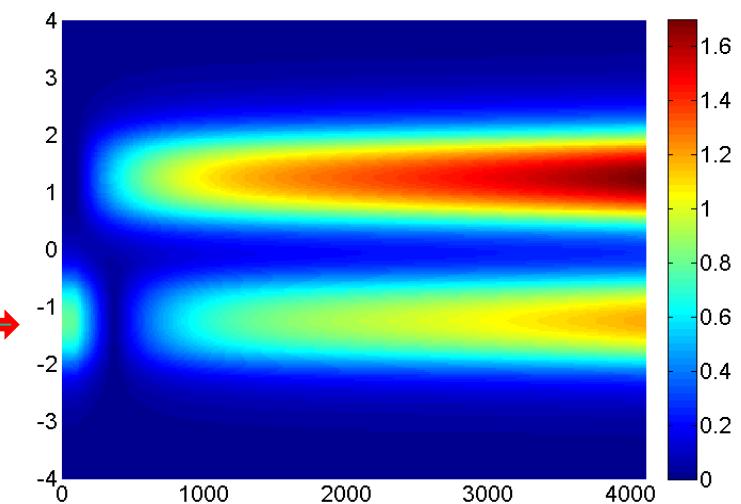
$$\varepsilon''_{g,loss} = 0.0105, \quad \varepsilon''_{g,gain} = -0.0065, \quad \varepsilon''_b = 0.002, \quad \varepsilon''_{g,loss} = 0.0115, \quad \varepsilon''_{g,gain} = -0.0075$$

Below critical point



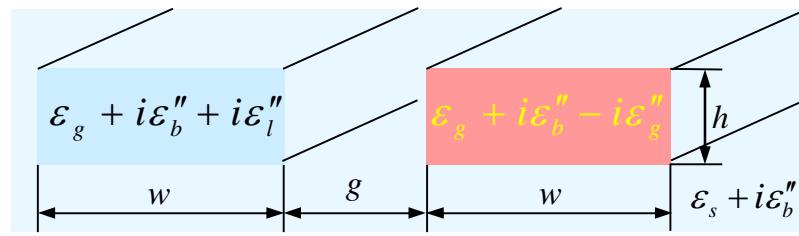
Above critical point  
gain channel  
loss channel

gain channel  
loss channel



# COUPLED WAVEGUIDES WITH LOSS/GAIN

Un/balanced loss/gain:



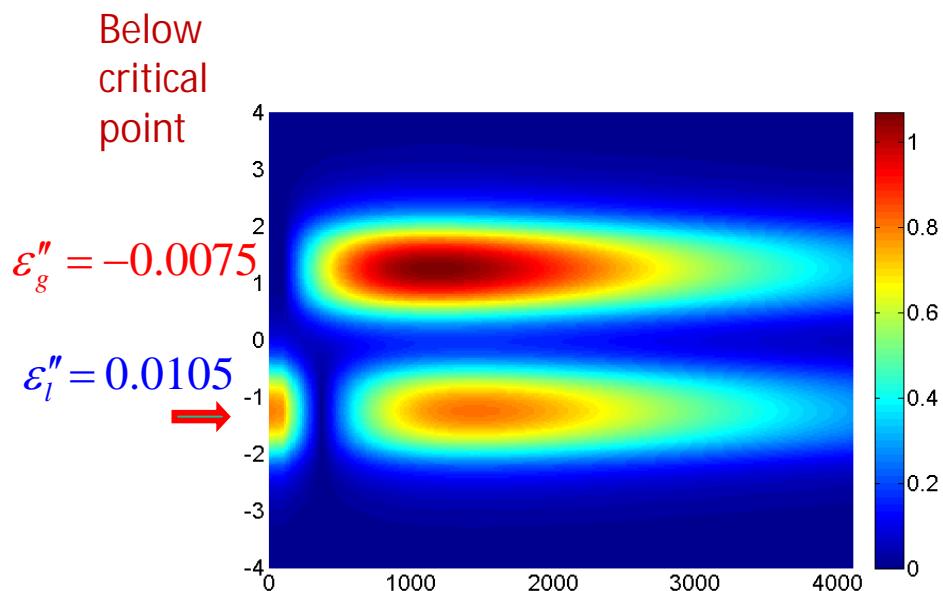
$$\epsilon_s = 10.89, \quad \epsilon_g = 11.56$$

$$w = 1.5 \text{ } \mu\text{m}, \quad h = 0.75 \text{ } \mu\text{m}, \\ g = 1 \text{ } \mu\text{m}, \quad \lambda = 1.55 \text{ } \mu\text{m}.$$

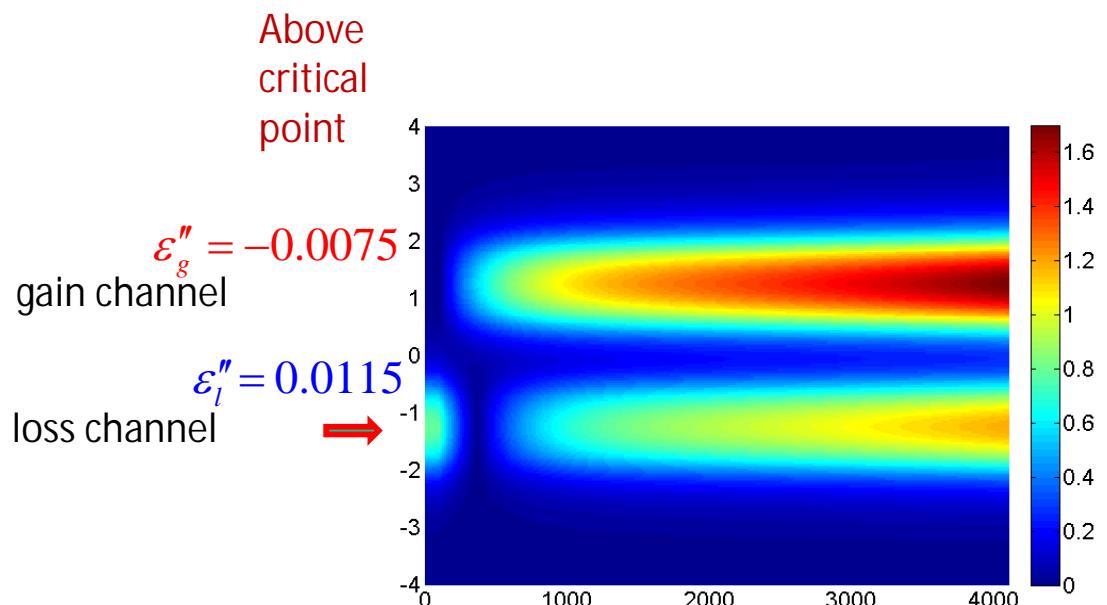
Structure with *uniform background loss*,  $\epsilon_b'' = 0.002$

Output *power increase* (from both waveguides!) by *increasing loss* of the lossy channel:

Lower loss, "subcritical" regime



Higher loss, "supercritical" regime



A. Guo et al., "Observation of PT-Symmetry Breaking in Complex Optical Potentials," *Physical Review Letters*, vol. 103, no. 9, pp. 093902-1-4, 2009.

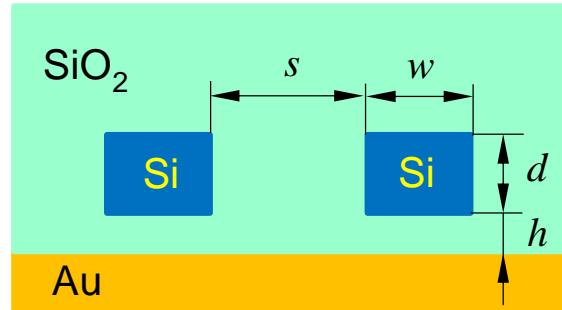
# PLASMONIC LOSS/GAIN STRUCTURES

A hypothetic “canonic” (balanced) plasmonic loss/gain structure:

Hybrid dielectric-plasmonic slot waveguide directional coupler *with “tunable metal”*

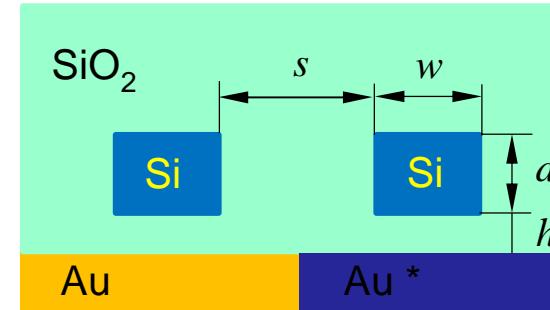
“Passive” structure:

$$\varepsilon(-x, y) = \varepsilon^*(x, y)$$

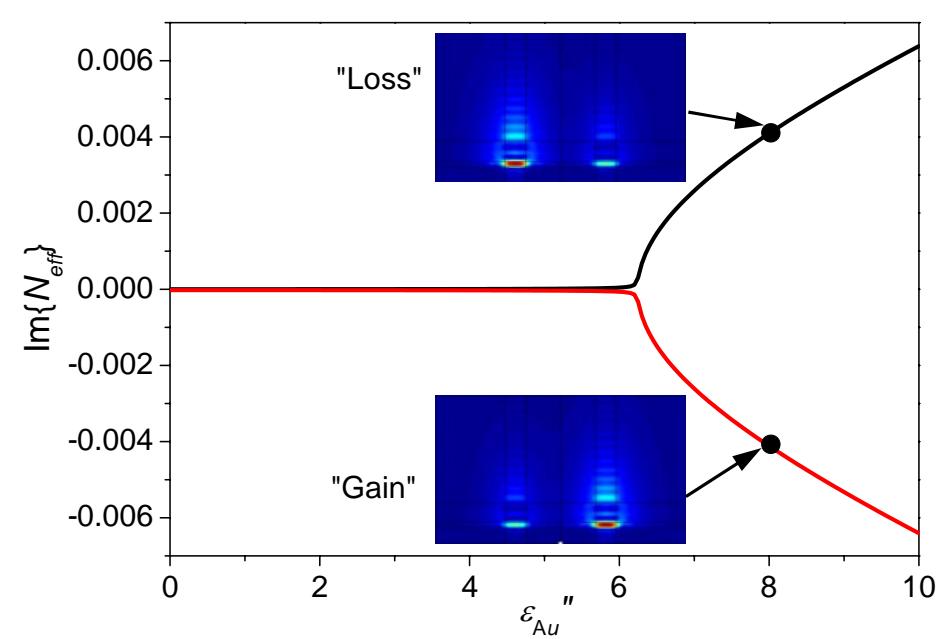
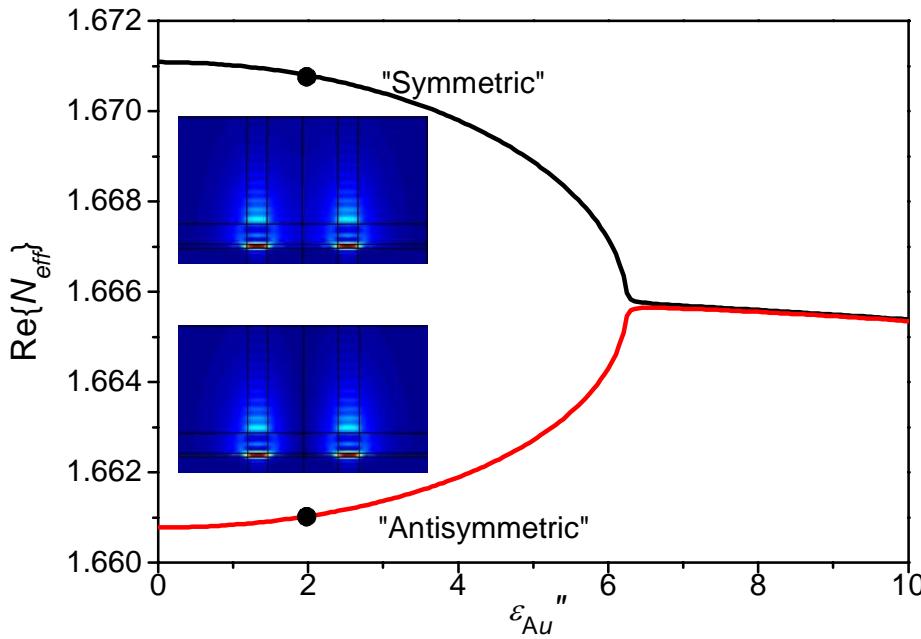


$$\begin{aligned}w &= 300 \text{ nm}, \\d &= 120 \text{ nm}, \\h &= 30 \text{ nm}, \\s &= 1000 \text{ nm}, \\\lambda &= 1.55 \mu\text{m}\end{aligned}$$

Au with tunable loss / Au\* with tunable gain



$$\varepsilon_{Au^*} = \varepsilon_{Au}^*$$

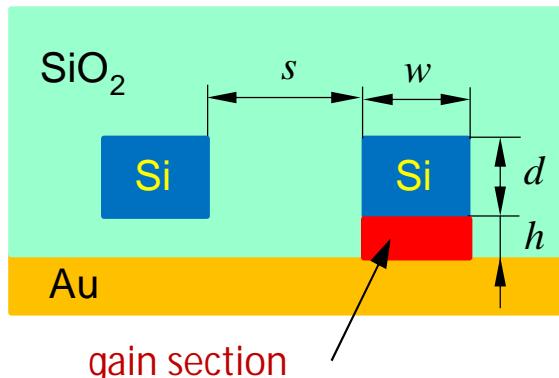


H. Benisty *et al.*, "Implementation of PT symmetric devices using plasmonics: principle and applications," *Optics Express*, vol. 19, pp. 18004-18019, 2011.

# PLASMONIC LOSS/GAIN STRUCTURES

A more realistic model of an *unbalanced* plasmonic loss/gain structure:

Hybrid dielectric-plasmonic slot waveguide directional coupler **with gain section**

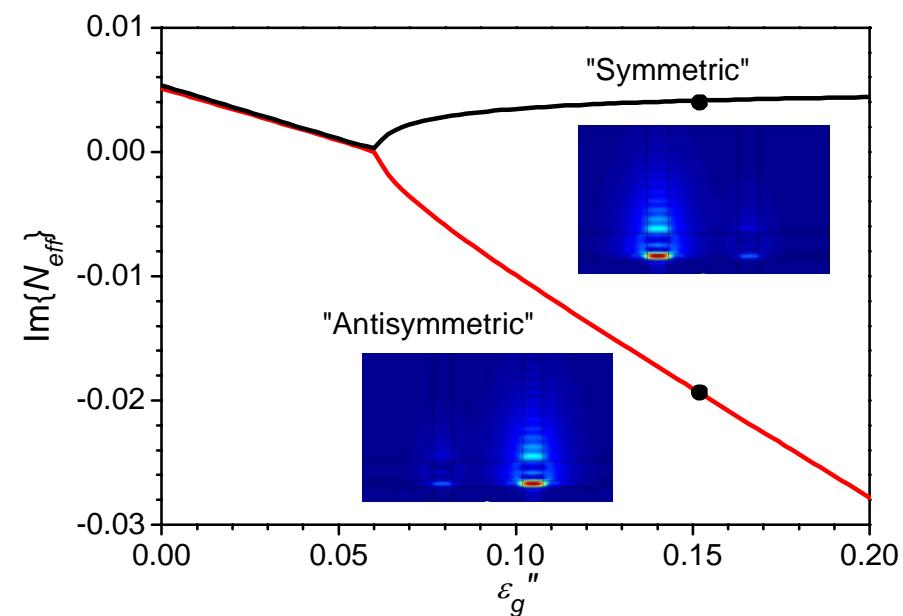
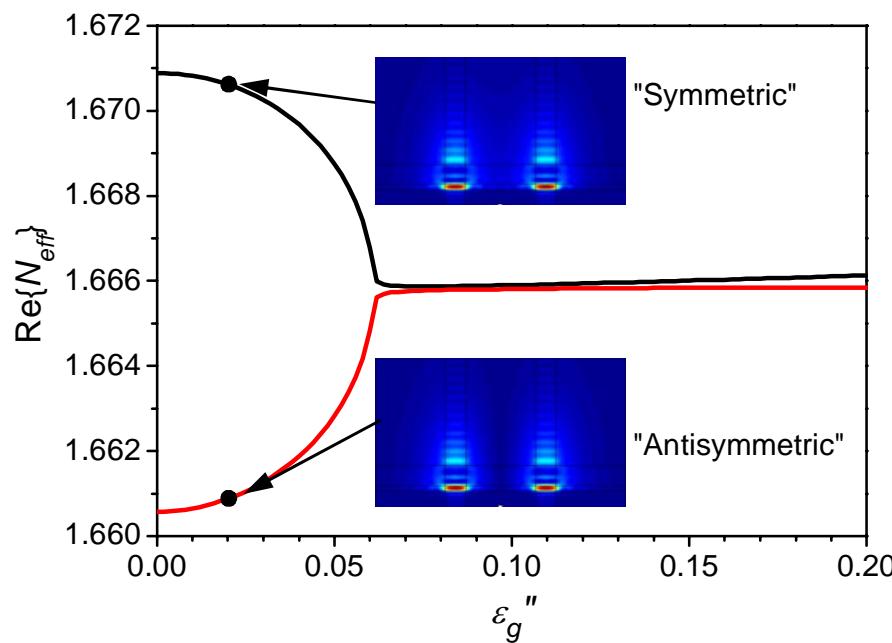


$$\begin{aligned}w &= 300 \text{ nm}, \\d &= 120 \text{ nm}, \\h &= 30 \text{ nm}, \\s &= 1000 \text{ nm}\end{aligned}$$

$$\varepsilon(-x, y) \neq \varepsilon^*(x, y)$$

Only gain ( $\varepsilon_g''$ ) in the gain section is now tuned:

$$\varepsilon_{\text{gain}} = \varepsilon_{\text{SiO}_2} - i\varepsilon_g''$$

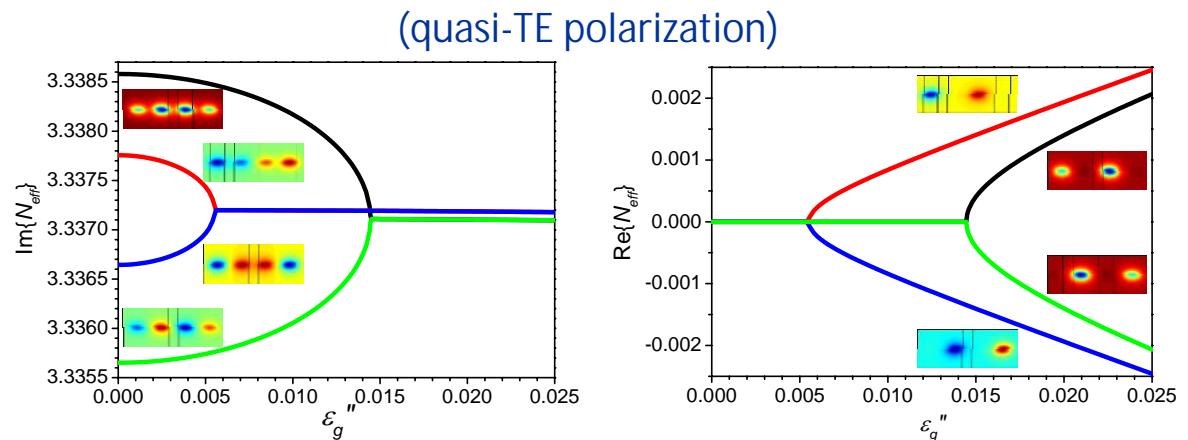
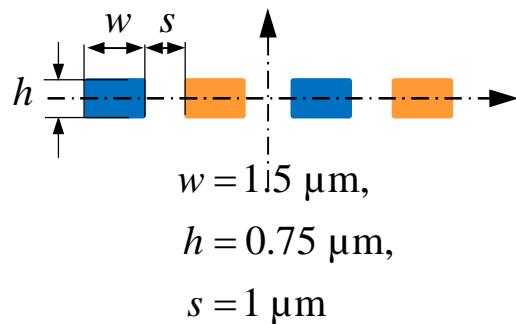


# MORE COMPLEX GAIN-LOSS STRUCTURES

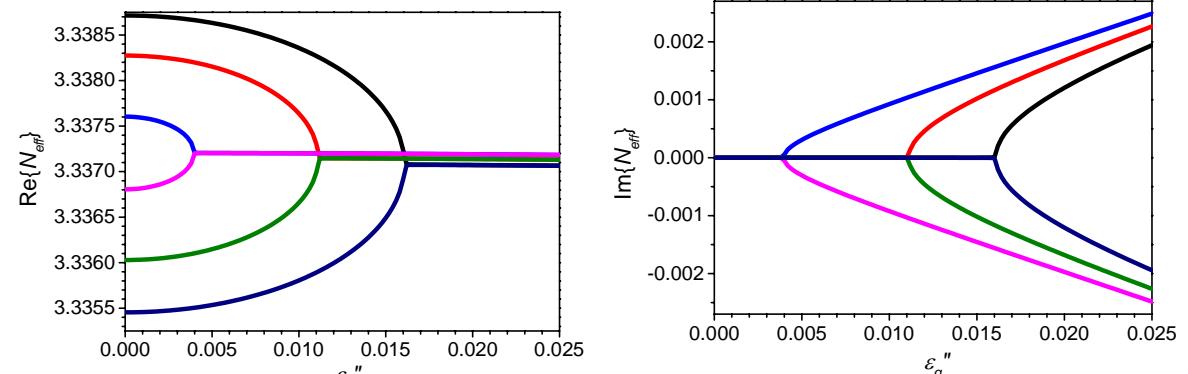
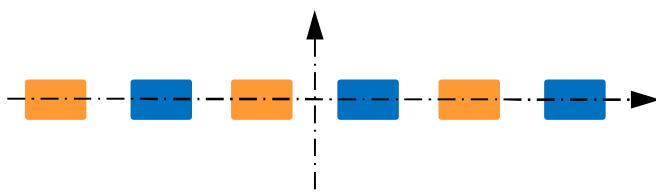
Linear arrays of coupled waveguides with loss and gain

$$\varepsilon(-x, y) = \varepsilon^*(x, y)$$

4 coupled channel waveguides

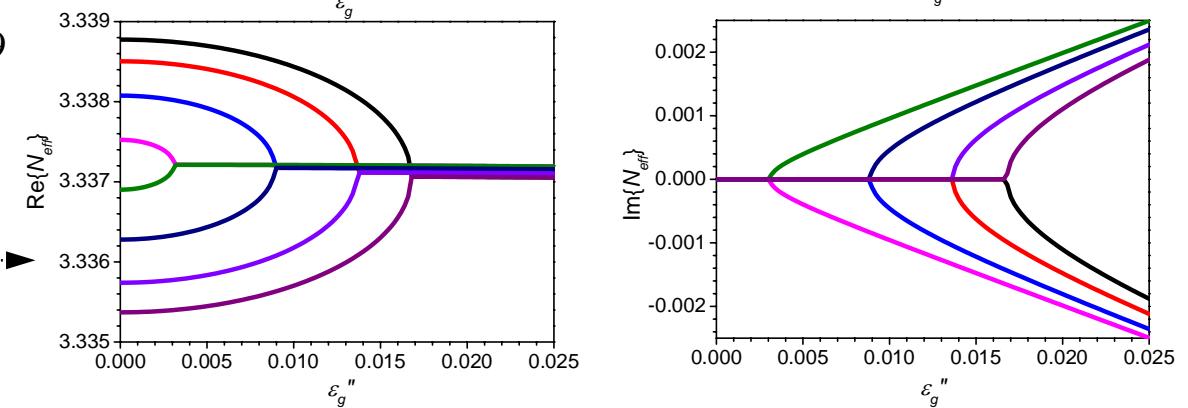
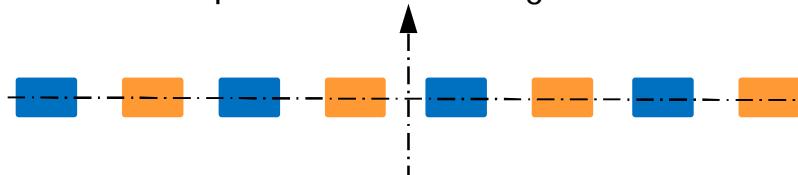


6 coupled channel waveguides



$$\varepsilon_{g1} = 11.56 + i\varepsilon_g'', \quad \varepsilon_{g2} = 11.56 - i\varepsilon_g'', \quad \varepsilon_s = 10.89$$

8 coupled channel waveguides

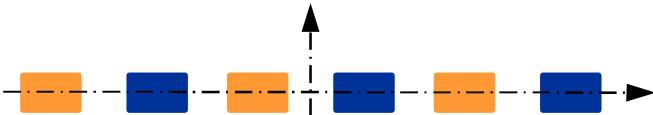


# LINEAR ARRAY WITH UNBALANCED LOSS/GAIN

Coupled waveguides with *fixed loss and variable gain*

$$\varepsilon(-x, y) \neq \varepsilon^*(x, y)$$

6 coupled channel waveguides

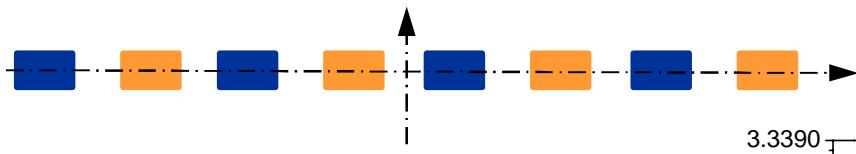


$$\varepsilon_{g1} = 11.56 + 0.011i,$$

$$\varepsilon_{g2} = 11.56 - i\varepsilon_g'',$$

$$\varepsilon_s = 10.89$$

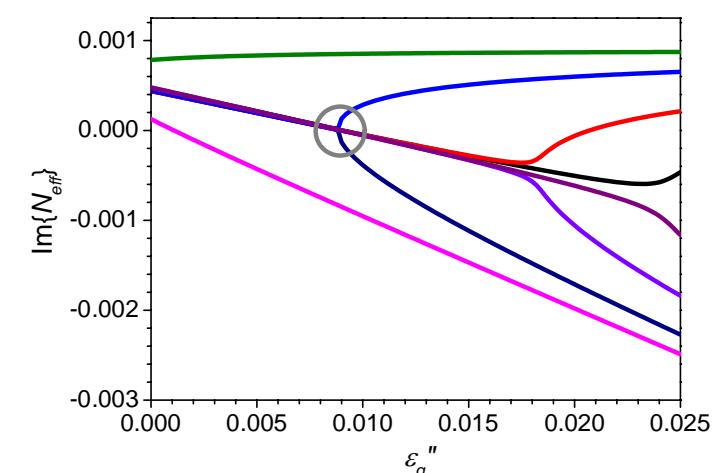
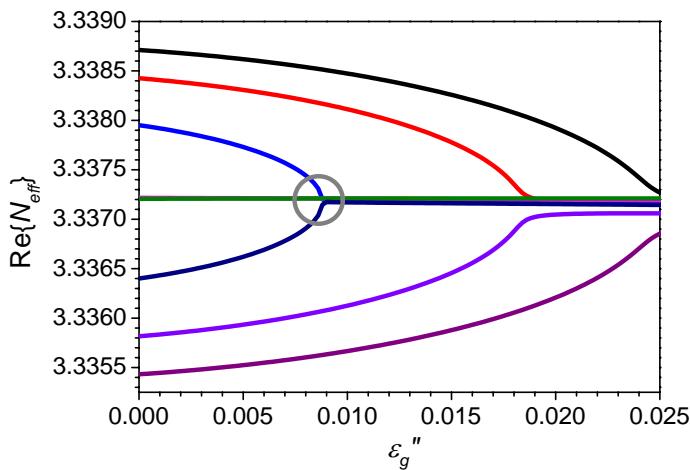
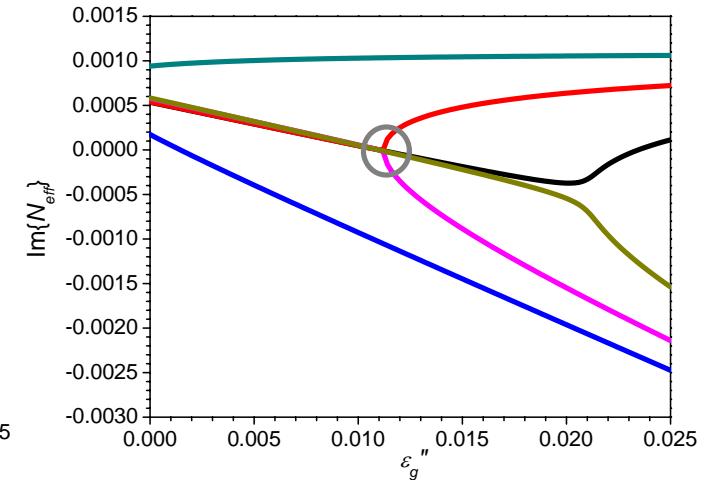
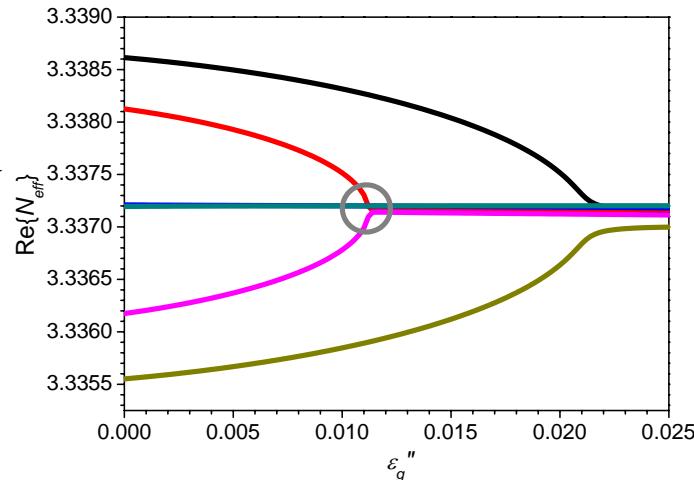
8 coupled channel waveguides



$$\varepsilon_{g1} = 11.56 + 0.009i,$$

$$\varepsilon_{g2} = 11.56 - i\varepsilon_g'',$$

$$\varepsilon_s = 10.89$$



“Switching” by pure gain modulation is feasible also in loss/gain waveguide arrays

# MORE COMPLEX GAIN-LOSS STRUCTURES

"Circular" arrays of coupled waveguides with loss and gain

4 waveguides

$$w = 1 \mu\text{m}$$

$$r = 1.5w$$

$$\epsilon_{g1} = 11.56 + i\epsilon_g'', \quad \epsilon_{g2} = 11.56 - i\epsilon_g''$$

$$\epsilon_s = 10.89$$

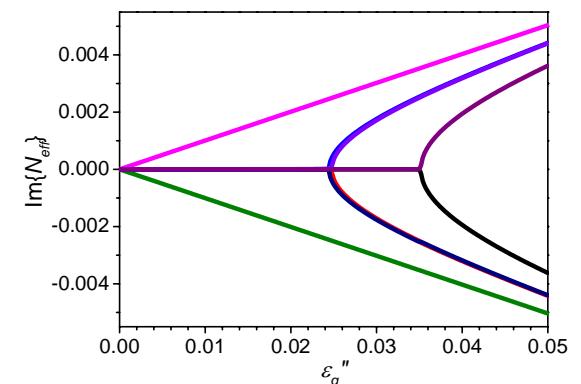
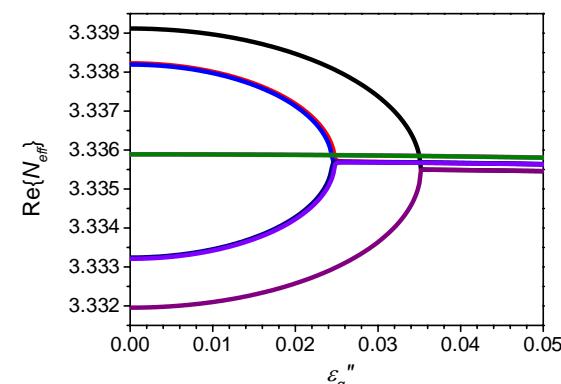
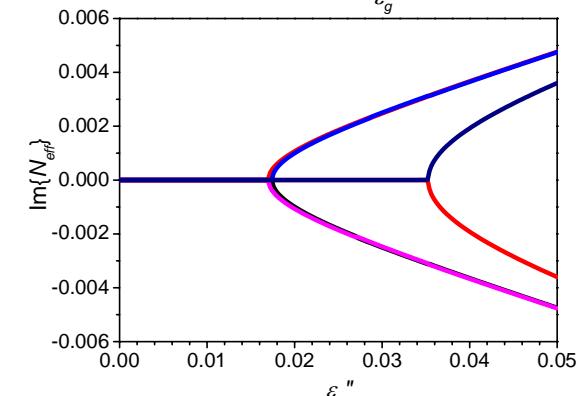
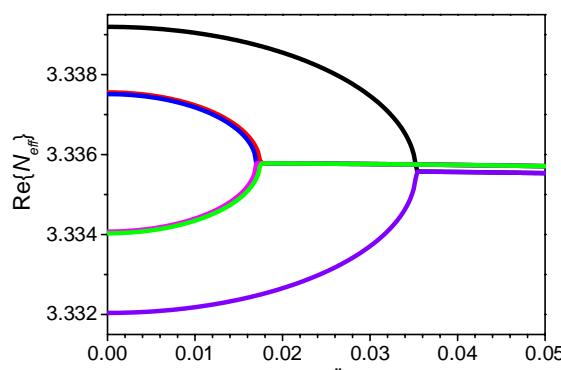
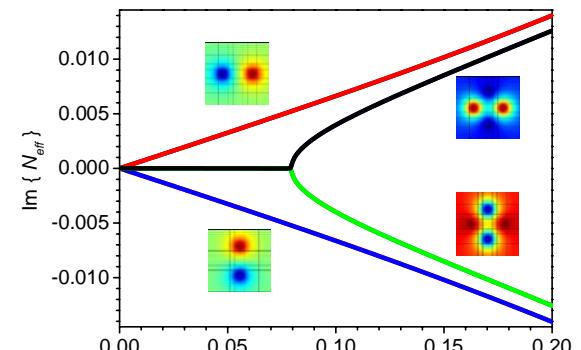
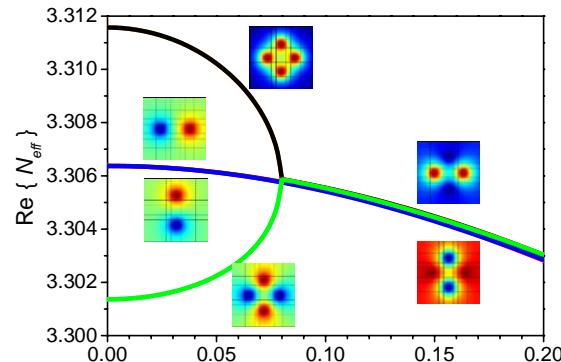
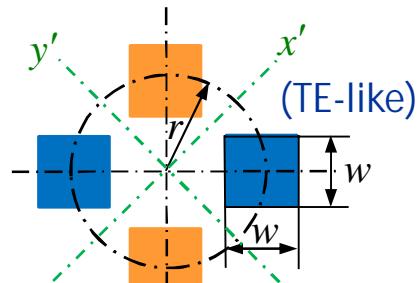
6 waveguides

$$r = 2w$$

8 waveguides

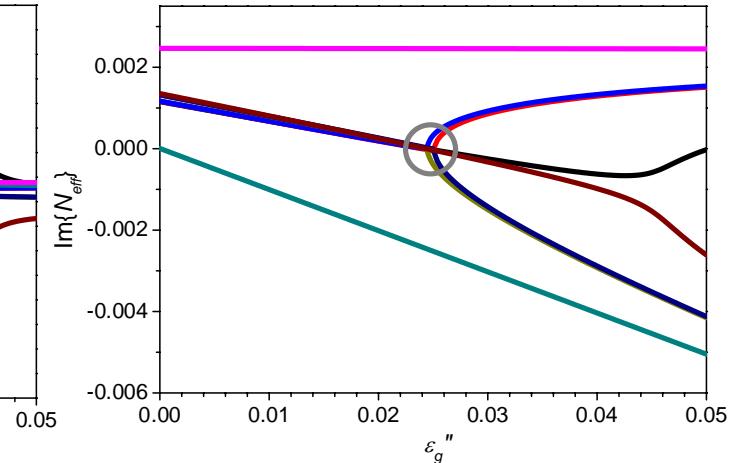
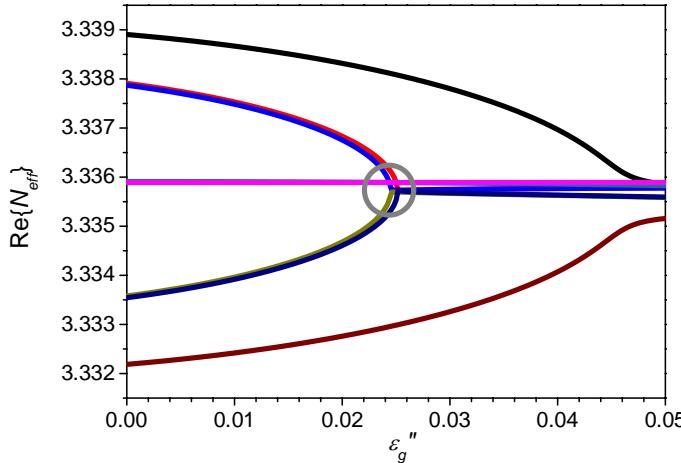
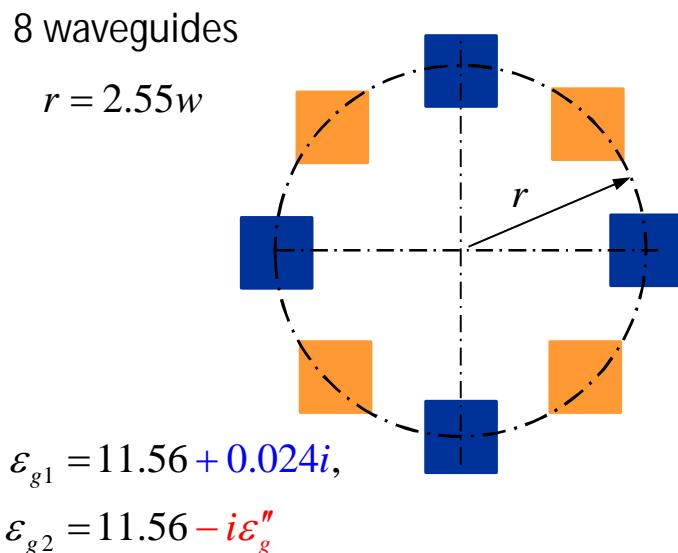
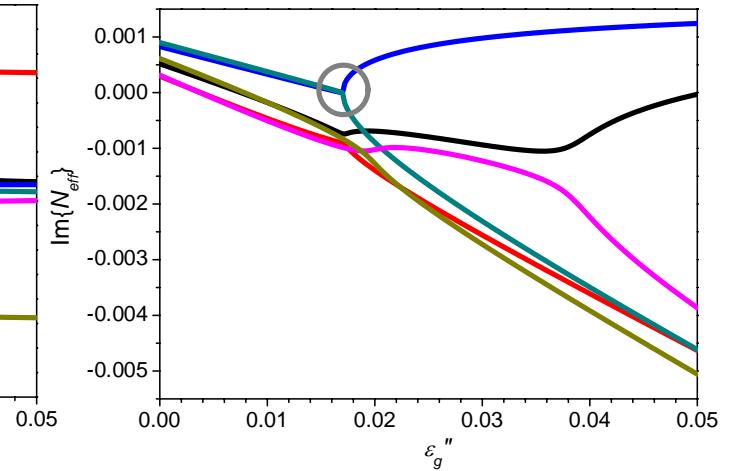
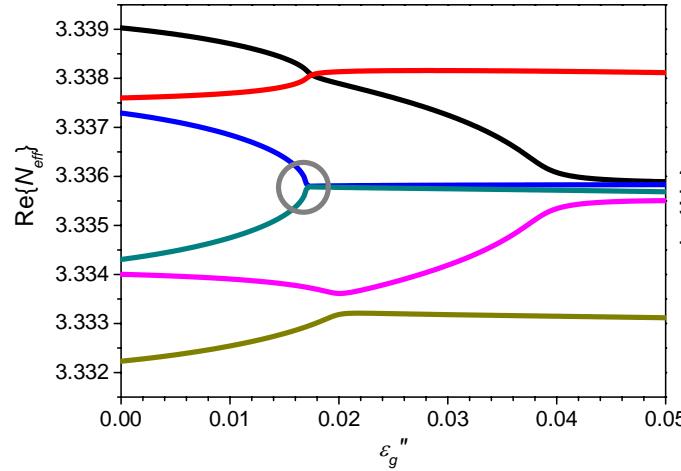
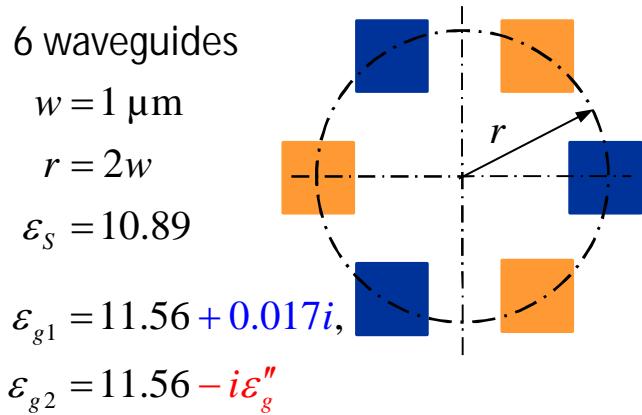
$$r = 2.55w$$

$$\begin{aligned} \epsilon(-x', y') &= \epsilon^*(x', y') \\ \epsilon(x', -y') &= \epsilon^*(x', y') \end{aligned}$$



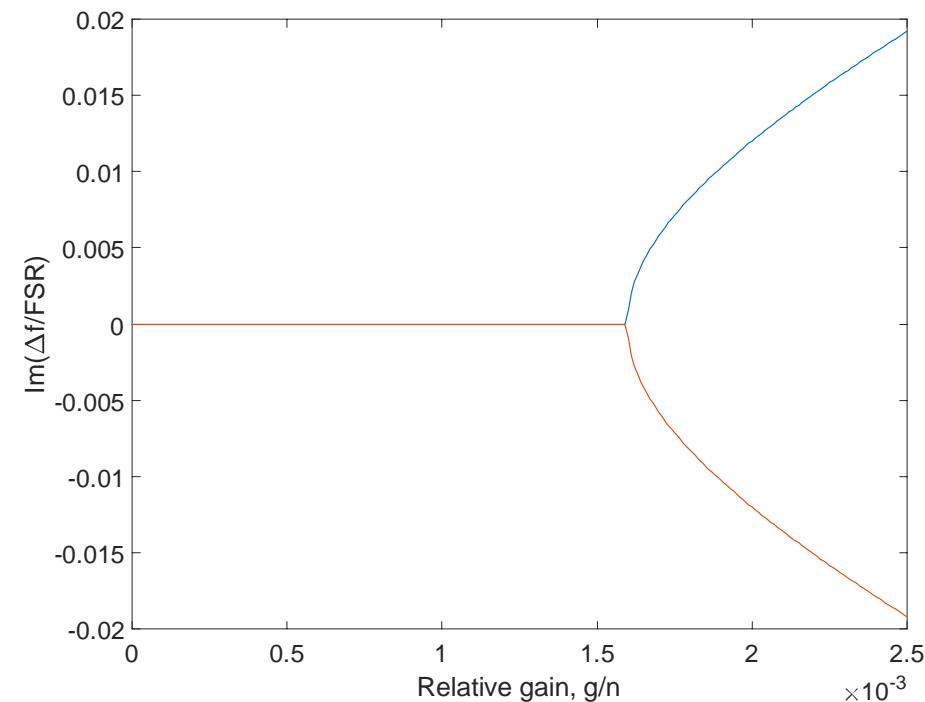
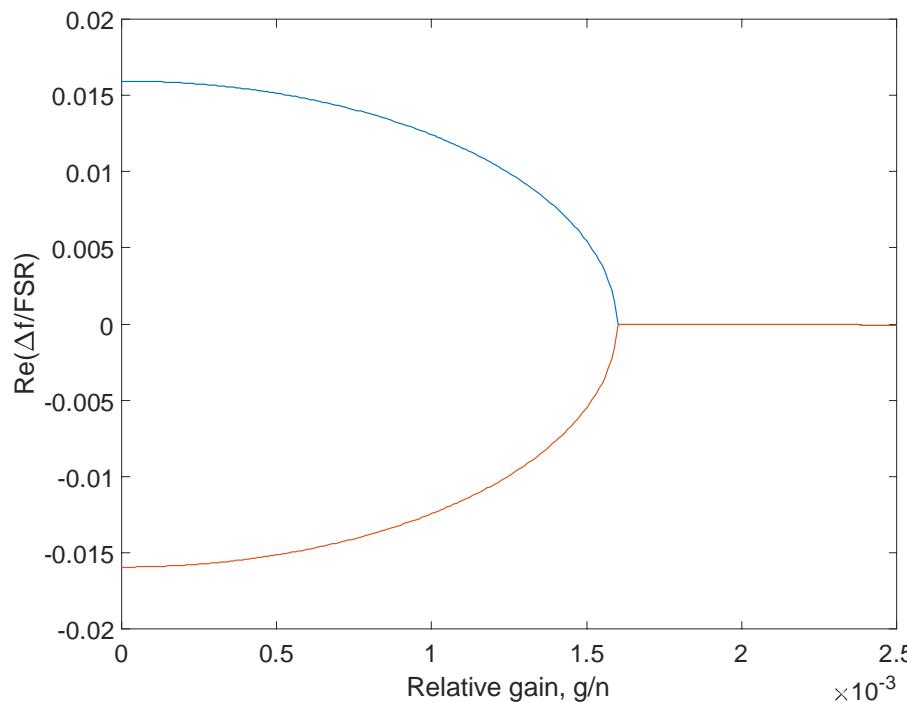
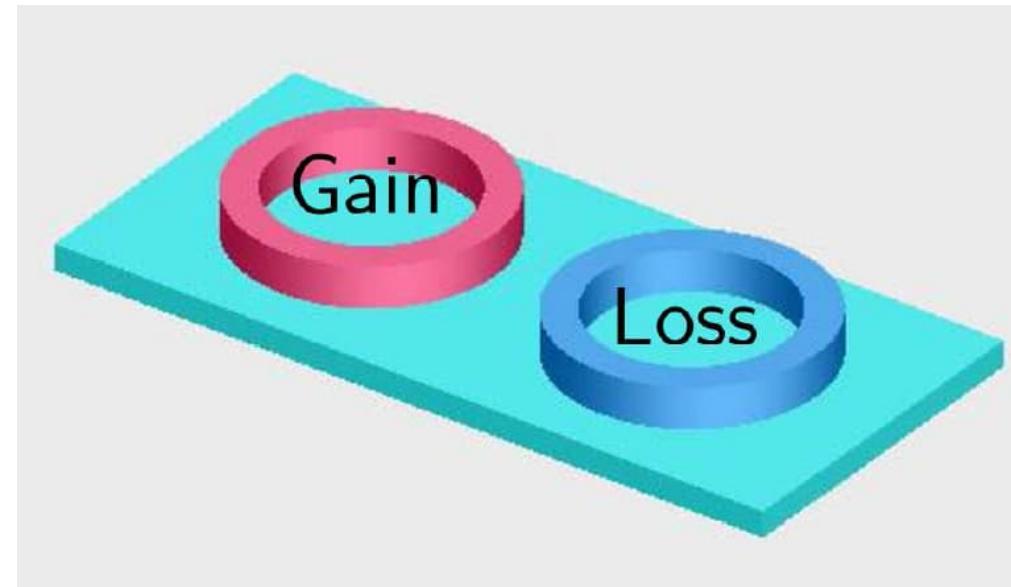
# CIRCULAR ARRAYS WITH UNBALANCED LOSS/GAIN

Coupled waveguides with *fixed loss* and *variable gain*



“Switching” by pure gain modulation is feasible also in loss/gain waveguide arrays

# RESONANT FREQUENCIES OF A PAIR OF COUPLED *PT*-SYMMETRIC RING RESONATORS



# SOME RELEVANT REFERENCES

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12. D. Chatzidimitriou, E. E. Kriezis, "Optical switching through graphene-induced exceptional points", JOSA B vol. 35, pp. 1525-1535, 2018
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